



TAS5342

SLAS557B-SEPTEMBER 2007-REVISED OCTOBER 2007

## **100 W STEREO DIGITAL AMPLIFIER POWER STAGE**

## **FEATURES**

- Total Power Output (Bridge Tied Load)
  - =  $2 \times 100$  W at 10% THD+N Into 4  $\Omega$
  - $2\times80$  W at 10% THD+N Into 6  $\Omega$
  - 2  $\times$  65 W at 10% THD+N Into 8 Ω
  - Total Power Output (Single Ended)
  - =  $4 \times 40$  W at 10% THD+N Into 3  $\Omega$
  - $4 \times 30$  W at 10% THD+N Into 4  $\Omega$
- Total Power Output (Parallel Mode)
- 1  $\times$  200 W at 10% THD+N Into 2  $\Omega$
- 1 × 160 W at 10% THD+N Into 3 Ω
- >110 dB SNR (A-Weighted With TAS5518 Modulator)
- <0.1% THD+N (1 W, 1 kHz)
- Supports PWM Frame Rates of 192 kHz to 432 kHz
- Resistor-Programmable Current Limit
- Integrated Self-Protection Circuitry, Including:
  - Under Voltage Protection
  - Overtemperature Warning and Error
  - Overload Protection
  - Short-Circuit Protection
  - PWM Activity Detector
- Standalone Protection Recovery
- Power-On Reset (POR) to Eliminate System Power-Supply Sequencing
- High-Efficiency Power Stage (>90%) With 80-mΩ Output MOSFETs
- Thermally Enhanced Package 44-Pin HTSSOP (DDV)
- Error Reporting, 3.3-V and 5.0-V Compliant
- EMI Compliant When Used With Recommended System Design

## **APPLICATIONS**

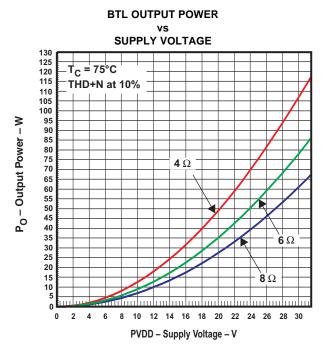
- Mini/Micro Audio System
- DVD Receiver
- Home Theater

## DESCRIPTION

The TAS5342 is a high-performance, integrated stereo digital amplifier power stage designed to drive a 4- $\Omega$  bridge-tied load (BTL) at up to 100 W per channel with low harmonic distortion, low integrated noise, and low idle current.

The TAS5342 has a complete protection system integrated on-chip, safeguarding the device against a wide range of fault conditions that could damage the system. These protection features are short-circuit protection, over-current protection, under voltage protection, over temperature protection, and a loss of PWM signal (PWM activity detector).

A power-on-reset (POR) circuit is used to eliminate power-supply sequencing that is required for most power-stage designs.



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## TAS5342



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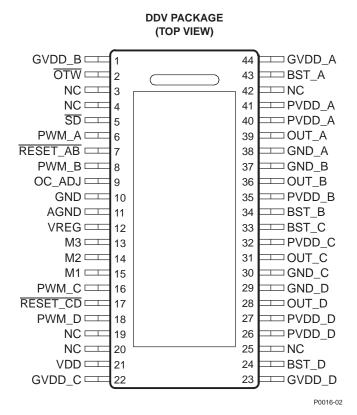
These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

### **GENERAL INFORMATION**

## **Terminal Assignment**

The TAS5342 is available in a thermally enhanced package 44-pin HTSSOP PowerPad™ package (DDV)

This package contains a thermal pad that is located on the top side of the device for convenient thermal coupling to the heatsink.



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## **Protection MODE Selection Pins**

MODE PINS PWM Input<sup>(1)</sup> Mode Name Description М3 M2 M1 BTL mode 1 0 0 0 2N All protection systems enabled 0 0 1 BTL mode 2 2N Latching shudown on, PWM activity detector and OLP disabled 0 1 0 BTL mode 3 1N All protection systems enabled 1N / 2N<sup>(2)</sup> 0 1 1 PBTL mode All protection systems enabled 1 0 0 SE mode 1 1N All protection systems enabled<sup>(3)</sup> 0 SE mode 2 1N Latching shudown on, PWM activity detector and OLP disabled<sup>(3)</sup> 1 1 0 1 1 Reserved

Protection modes are selected by shorting M1, M2, and M3 to VREG or GND.

The 1N and 2N naming convention is used to indicate the number of PWM lines to the power stage per channel in a specific mode. (1) PWM D is used to select between the 1N and 2N interface in PBTL mode (Low = 1N; High = 2N). PWM D is internally pulled low in (2)

PBTL mode. PWM\_A is used as the PWM input in 1N mode and PWM\_A and PWM\_B are used as inputs for the 2N mode.

(3) PPSC detection system disabled.

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#### Package Heat Dissipation Ratings<sup>(1)</sup>

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PARAMETER	TAS5342DDV
R <sub>0JC</sub> (°C/W)—2 BTL or 4 SE channels	1.3
R <sub>0JC</sub> (°C/W)—1 BTL or 2 SE channel(s)	2.6
R <sub>0JC</sub> (°C/W)—1 SE channel	5.0
Power Pad area <sup>(2)</sup>	36 mm <sup>2</sup>

(1) JC is junction-to-case, CH is case-to-heatsink.

R<sub>eCH</sub> is an important consideration. Assume a 2-mil thickness of high performance grease with a thermal conductivity at 2.5W/m-K (2) between the pad area and the heat sink. The  $R_{\theta CH}$  with this condition is 0.6°C/W for the DDV package.



#### **ABSOLUTE MAXIMUM RATINGS**

over operating free-air temperature range unless otherwise noted (1)

TAS5342	
VDD to AGND	–0.3 V to 13.2 V
GVDD_X to AGND	–0.3 V to 13.2 V
PVDD_X to GND_X (2)	–0.3 V to 53 V
OUT_X to GND_X <sup>(2)</sup>	–0.3 V to 53 V
BST_X to GND_X <sup>(2)</sup>	-0.3 V to 66.2 V
BST_X to GVDD_X <sup>(2)</sup>	–0.3 V to 53 V
VREG to AGND	–0.3 V to 4.2 V
GND_X to GND	–0.3 V to 0.3 V
GND_X to AGND	–0.3 V to 0.3 V
GND to AGND	–0.3 V to 0.3 V
PWM_X, OC_ADJ, M1, M2, M3 to AGND	–0.3 V to 4.2 V
RESET_X, SD, OTW to AGND	–0.3 V to 7 V
Maximum continuous sink current (SD, OTW)	9 mA
Maximum operating junction temperature range, T <sub>J</sub>	0°C to 125°C
Storage temperature	–40°C to 125°C
Lead temperature, 1,6 mm (1/16 inch) from case for 10 seconds	260°C
Minimum pulse duration, low	30 ns

(1) Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

(2) These voltages represent the dc voltage + peak ac waveform measured at the terminal of the device in all conditions.

#### ORDERING INFORMATION<sup>(1)</sup>

T <sub>A</sub>	PACKAGE <sup>(1)</sup>	DESCRIPTION
0°C to 70°C	TAS5342DDV	44-pin HTSSOP

(1) For the most current package and ordering information, see the Package Option Addendum at the end of this document, or see the TI website at www.ti.com.



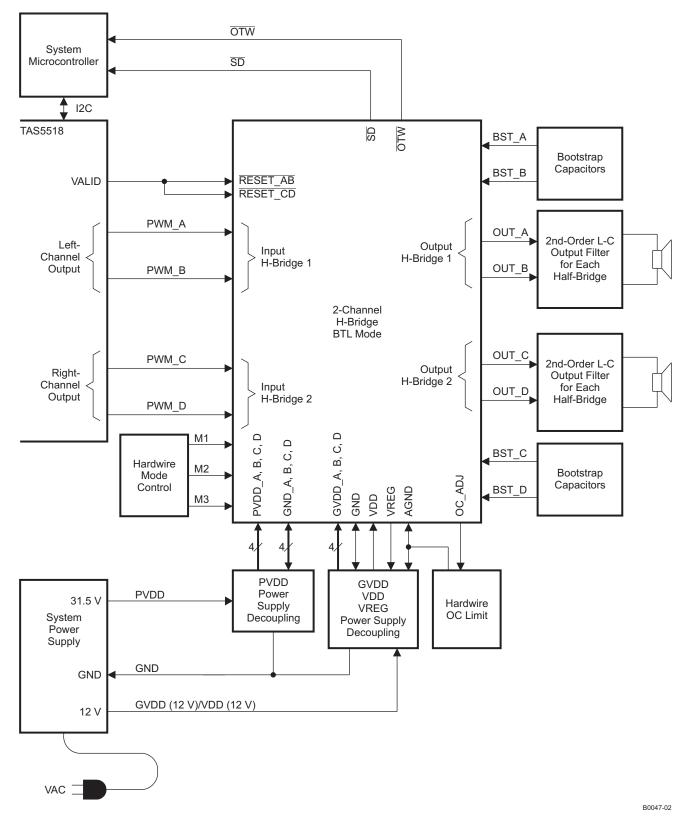
## **Terminal Functions**

TE	TERMINAL		DECODIDITION (	
NAME	DDV NO.	FUNCTION <sup>(1)</sup>	DESCRIPTION	
AGND	11	Р	Analog ground	
BST_A	43	Р	Bootstrap pin, A-Side	
BST_B	34	Р	Bootstrap pin, B-Side	
BST_C	33	Р	Bootstrap pin, C-Side	
BST_D	24	Р	Bootstrap pin, D-Side	
GND	10	Р	Ground	
GND_A	38	Р	Power ground for half-bridge A	
GND_B	37	Р	Power ground for half-bridge B	
GND_C	30	Р	Power ground for half-bridge C	
GND_D	29	Р	Power ground for half-bridge D	
GVDD_A	44	Р	Gate-drive voltage supply; A-Side	
GVDD_B	1	Р	Gate-drive voltage supply; B-Side	
GVDD_C	22	Р	Gate-drive voltage supply; C-Side	
GVDD_D	23	Р	Gate-drive voltage supply; D-Side	
M1	15	I	Mode selection pin (LSB)	
M2	14	I	Mode selection pin	
M3	13	I	Mode selection pin (MSB)	
NC	3, 4, 19, 20, 25, 42	_	No connect. Pins may be grounded.	
OC_ADJ	9	0	Analog overcurrent programming pin	
OTW	2	0	Overtemperature warning signal, open-drain, active-low	
OUT_A	39	0	Output, half-bridge A	
OUT_B	36	0	Output, half-bridge B	
OUT_C	31	0	Output, half-bridge C	
OUT_D	28	0	Output, half-bridge D	
PVDD_A	40, 41	Р	Power supply input for half-bridge A	
PVDD_B	35	Р	Power supply input for half-bridge B	
PVDD_C	32	Р	Power supply input for half-bridge C	
PVDD_D	26, 27	Р	Power supply input for half-bridge D	
PWM_A	6	Ι	PWM Input signal for half-bridge A	
PWM_B	8	Ι	PWM Input signal for half-bridge B	
PWM_C	16	I	PWM Input signal for half-bridge C	
PWM_D	18	I	PWM Input signal for half-bridge D	
RESET_AB	7	I	Reset signal for half-bridge A and half-bridge B, active-low	
RESET_CD	17	I	Reset signal for half-bridge C and half-bridge D, active-low	
SD	5	0	Shutdown signal, open-drain, active-low	
VDD	21	Р	Input power supply	
VREG	12	Р	Internal voltage regulator	

(1) I = input, O = output, P = power

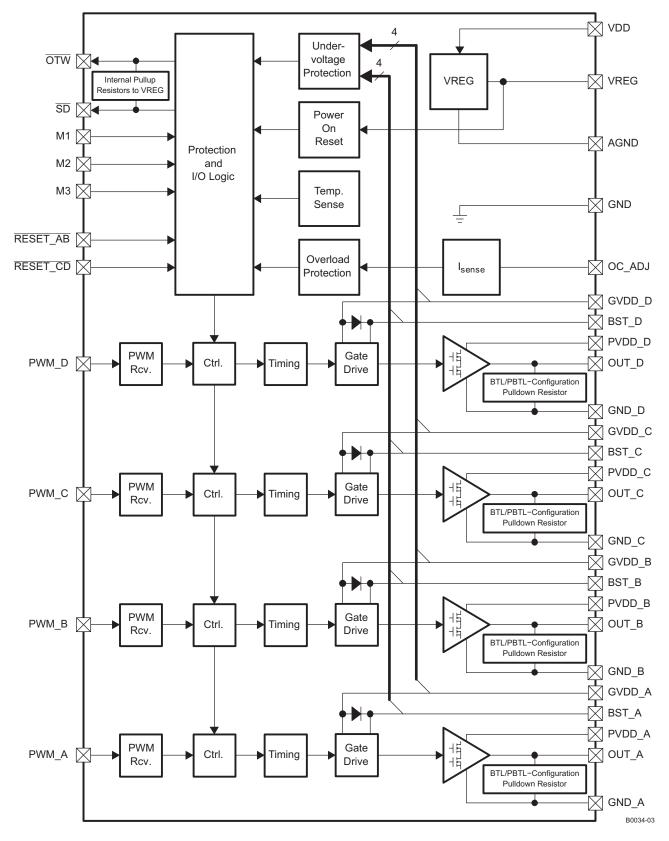


## **TYPICAL SYSTEM BLOCK DIAGRAM**





## FUNCTIONAL BLOCK DIAGRAM





## **RECOMMENDED OPERATING CONDITIONS**

			MIN	TYP	MAX	UNIT
PVDD_X	Half-bridge supply voltage		0	31.5	34	V
GVDD_X	Supply voltage for logic regulators and gate-drive circuitry		10.8	12	13.2	V
VDD	Digital regulator supply voltage		10.8	12	13.2	V
R <sub>L</sub> (BTL)	Resistive load impedance (no Cycle-by Cycle		3	4		
R <sub>L</sub> (SE)	current control), recommended demodulation		2.25	3		Ω
R <sub>L</sub> (PBTL)	filter		1.5	2		
L <sub>Output</sub> (BTL)			5	10		
L <sub>Output</sub> (SE)	Output-filter inductance	Minimum output inductance under short-circuit condition	5	10		μH
L <sub>Output</sub> (PBTL)			5	10		
f <sub>S</sub>	PWM frame rate		192	384	432	kHz
t <sub>LOW</sub>	Minimum low-state pulse duration per PWM Frame, noise shaper enabled		30			nS
C <sub>PVDD</sub>	PVDD close decoupling capacitors			0.1		μF
C <sub>BST</sub>	Bootstrap capacitor, selected value supports PWM frame rates from 192 kHz to 432 kHz			33		nF
R <sub>OC</sub>	Over-current programming resistor	Resistor tolerance = 5%	27	27	47	kΩ
R <sub>EXT-PULLUP</sub>	External pull-up resistor to +3.3V to +5.0V for SD or OTW		3.3	4.7		kΩ
TJ	Junction temperature		0		125	°C

## AUDIO SPECIFICATIONS (BTL)

Audio performance is recorded as a chipset consisting of a TAS5518 pwm processor (modulation index limited to 97.7%) and a TAS5342 power stage. PCB and system configuraton are in accordance with recommended guidelines. Audio frequency = 1 kHz, PVDD\_x = 31.5 V, GVDD\_x = 12 V, R<sub>L</sub> = 4  $\Omega$ , f<sub>S</sub> = 384 kHz, R<sub>OC</sub> = 27 k $\Omega$ , T<sub>C</sub> = 75°C, Output Filter: L<sub>DEM</sub> = 10  $\mu$ H, C<sub>DEM</sub> = 470 nF, unless otherwise noted.

PARAMETER		TEST CONDITIONS	TAS5342			UNIT
	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
		$R_L$ = 4 Ω, 10% THD+N, clipped input signal		100		
P <sub>OMAX</sub>	Maximum Power Output	$R_L$ = 6 Ω, 10% THD+N, clipped input signal		80		
		R <sub>L</sub> = 8 Ω, 10% THD+N, clipped input signal		65		W
	P <sub>O</sub> Unclipped Power Output	$R_L$ = 4 $\Omega$ , 0 dBFS, unclipped input signal		80		
Po		$R_L$ = 6 $\Omega$ , 0 dBFS, unclipped input signal		64		
		$R_L$ = 8 $\Omega$ , 0 dBFS, unclipped input signal		50		
THD+N	Total harmonic distortion + noise	0 dBFS; AES17 filter		0.2%		
I DHN	Total harmonic distortion + hoise	1 W; AES17 filter		0.09%		
V <sub>n</sub>	Output integrated noise	A-weighted, AES17 filter, Auto mute disabled		45		μV
SNR	Signal-to-noise ratio <sup>(1)</sup>	A-weighted, AES17 filter, Auto mute disabled		110		dB
DNR	Dynamic range	A-weighted, input level = –60 dBFS, AES17 filter		110		dB
DC Offset	Output offset voltage			+/- 15		mV
P <sub>idle</sub>	Power dissipation due to idle losses (IPVDD_X)	$P_O = 0$ W, all halfbridges switching <sup>(2)</sup>		2		W

(1) SNR is calculated relative to 0-dBFS input level.

(2) Actual system idle losses are affected by core losses of output inductors.

## AUDIO SPECIFICATIONS (Single-Ended Output)

Audio performance is recorded as a chipset consisting of a TAS5086 pwm processor (modulation index limited to 97.7%) and a TAS5342 power stage. PCB and system configuraton are in accordance with recommended guidelines. Audio frequency = 1 kHz, PVDD\_x = 31.5 V, GVDD\_x = 12 V, R<sub>L</sub> = 4  $\Omega$ , f<sub>S</sub> = 384 kHz, R<sub>OC</sub> = 27 k $\Omega$ , T<sub>C</sub> = 75°C, Output Filter: L<sub>DEM</sub> = 20  $\mu$ H, C<sub>DEM</sub> = 1.0  $\mu$ F, unless otherwise noted.

PARAMETER		TEST CONDITIONS	T	UNIT		
	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
D	Movimum Dowor Output	R <sub>L</sub> = 3 Ω, 10% THD+N, clipped input signal		40		
P <sub>OMAX</sub>	Maximum Power Output	$R_L$ = 4 Ω, 10% THD+N, clipped input signal		30		W
D	Unclipped Power Output	$R_L$ = 3 $\Omega$ , 0 dBFS, unclipped input signal		30		vv
Po		$R_L$ = 4 $\Omega$ , 0 dBFS, unclipped input signal		20		
THD+N	Total harmonic distortion + noise	0 dBFS; AES17 filter		0.2%		
	Total Harmonic distortion + hoise	1 W; AES17 filter		0.09%		
V <sub>n</sub>	Output integrated noise	A-weighted, AES17 filter, Auto mute disabled		35		μV
SNR	Signal-to-noise ratio <sup>(1)</sup>	A-weighted, AES17 filter, Auto mute disabled		109		dB
DNR	Dynamic range	A-weighted, input level = –60 dBFS AES17 filter		109		dB
P <sub>idle</sub>	Power dissipation due to idle losses (IPVDD_X)	$P_O = 0$ W, all half bridges switching <sup>(2)</sup>		2		W

(1) SNR is calculated relative to 0-dBFS input level.

(2) Actual system idle losses are affected by core losses of output inductors.

## AUDIO SPECIFICATIONS (PBTL)

Audio performance is recorded as a chipset consisting of a TAS5518 pwm processor (modulation index limited to 97.7%) and a TAS5342 power stage. PCB and system configuraton are in accordance with recommended guidelines. Audio frequency = 1kHz, PVDD\_x = 31.5 V, GVDD\_x = 12 V, R<sub>L</sub> = 3  $\Omega$ , f<sub>S</sub> = 384 kHz, R<sub>OC</sub> = 27 k $\Omega$ , T<sub>C</sub> = 75°C, Output Filter: L<sub>DEM</sub> = 10  $\mu$ H, C<sub>DEM</sub> = 1.0 uF, unless otherwise noted.

PARAMETER		TEST CONDITIONS	TAS5342			
	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
P	Maximum Bawar Output	$R_L$ = 2 $\Omega$ , 10% THD+N, clipped input signal		200		
P <sub>OMAX</sub>	Maximum Power Output	$R_L$ = 3 $\Omega$ , 10% THD+N, clipped input signal		160		W
D	Unclipped Power Output	$R_L$ = 2 $\Omega$ , 0 dBFS, unclipped input signal		150		vv
Po		$R_L$ = 3 $\Omega$ , 0 dBFS, unclipped input signal		120		
THD+N	Total harmonic distortion + noise	0 dBFS; AES17 filter		0.2%		
	Total harmonic distortion + hoise	1 W; AES17 filter		0.09%		
V <sub>n</sub>	Output integrated noise	A-weighted, AES17 filter, Auto mute disabled		45		μV
SNR	Signal-to-noise ratio <sup>(1)</sup>	A-weighted, AES17 filter, Auto mute disabled		110		dB
DNR	Dynamic range	A-weighted, input level = -60 dBFS AES17 filter		110		dB
DC Offset	Outuput offset voltage			+/- 15		mV

(1) SNR is calculated relative to 0-dBFS input level.



### AUDIO SPECIFICATIONS (PBTL) (continued)

Audio performance is recorded as a chipset consisting of a TAS5518 pwm processor (modulation index limited to 97.7%) and a TAS5342 power stage. PCB and system configuraton are in accordance with recommended guidelines. Audio frequency = 1kHz, PVDD\_x = 31.5 V, GVDD\_x = 12 V, R<sub>L</sub> = 3  $\Omega$ , f<sub>s</sub> = 384 kHz, R<sub>oc</sub> = 27 k $\Omega$ , T<sub>c</sub> = 75°C, Output Filter: L<sub>DEM</sub> = 10  $\mu$ H, C<sub>DEM</sub> = 1.0 uF, unless otherwise noted.

PARAMETER		TEST CONDITIONS	TAS5342			UNIT
	FARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
P <sub>idle</sub>	Power dissipation due to idle losses (IPVDD_X)	P <sub>O</sub> = 0 W, all half bridges switching <sup>(2)</sup>		2		W

(2) Actual system idle losses are affected by core losses of output inductors.

### **ELECTRICAL CHARACTERISTICS**

PVDD\_x = 31.5 V, GVDD\_X = 12 V, VDD = 12 V, T<sub>C</sub> (Case temperature) = 25°C, f<sub>S</sub> = 384 kHz, unless otherwise specified.

DADAMETED			TAS5342			
	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Internal Voltage	e Regulator and Current Consumption					
VREG	Voltage regulator, only used as a reference node	VDD = 12 V	3	3.3	3.6	V
IVDD		Operating, 50% duty cycle		7.2	17	A
IVDD	VDD supply current	Idle, reset mode		5.54	11	mA
IGVDD X	Gate supply current per half-bridge	50% duty cycle		8	16	mA
IGVDD_X	Gate supply current per nall-blidge	Reset mode		1	1.8	ША
IPVDD_X	Half-bridge idle current	50% duty cycle, without output filter or load		16.6	25	mA
		Reset mode, no switching		465	558	μA
Output Stage M	IOSFETs					
R <sub>DSon,LS</sub>	Drain-to-source resistance, Low Side	$T_J$ = 25°C, excludes metallization resistance,		80	89	mΩ
R <sub>DSon,HS</sub>	Drain-to-source resistance, High Side	$T_J$ = 25°C, excludes metallization resistance,		80	89	mΩ
I/O Protection						
V <sub>uvp,G</sub>	Undervoltage protection limit, GVDD_X			9.5		V
V <sub>uvp,hyst</sub> <sup>(1)</sup>	Undervoltage protection limit, GVDD_X			250		mV
BST <sub>uvpF</sub>	Puts device into RESET when BST voltage falls below limit			5.85		V
BST <sub>uvpR</sub>	Brings device out of RESET when BST voltage rises above limit			7		V
OTW <sup>(1)</sup>	Overtemperature warning		115	125	135	°C
OTW <sub>HYST</sub> <sup>(1)</sup>	Temperature drop needed below OTW temp. for OTW to be inactive after the OTW event			25		°C
OTE <sup>(1)</sup>	Overtemperature error threshold		145	155	165	°C
OTE- OTW <sub>differential</sub> <sup>(1)</sup>	OTE - OTW differential, temperature delta between OTW and OTE			30		°C
OLPC	Overload protection counter	f <sub>S</sub> = 384 kHz		1.25		ms
l <sub>oc</sub>	Overcurrent limit protection	Resistor—programmable, high-end, R <sub>OC</sub> = 27 k $\Omega$ with 1 ms pulse		10.1		А
I <sub>OCT</sub>	Overcurrent response time			150		ns
t <sub>ACTIVITY</sub> DETECTOR	Time for PWM activity detector to activite when no PWM is present	Lack of transistion of any PWM input		13.2		μS

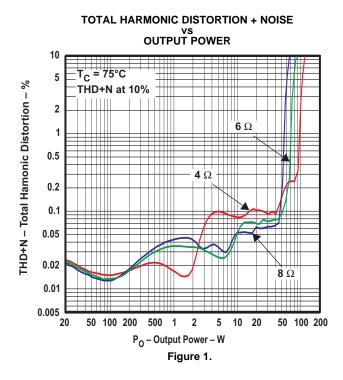


## **ELECTRICAL CHARACTERISTICS (continued)**

PVDD\_x = 31.5 V, GVDD\_X = 12 V, VDD = 12 V, T<sub>C</sub> (Case temperature) = 25°C, f<sub>S</sub> = 384 kHz, unless otherwise specified.

PARAMETER		TEST CONDITIONS	TAS5342			
		TEST CONDITIONS	MIN	TYP	MAX	UNIT
I <sub>PD</sub>	Output pulldown current of each half-bridge	Connected when RESET is active to provide bootstrap capacitor charge. Not used in SE mode.		3		mA
Static Digital	Specifications					
V <sub>IH</sub>	High-level input voltage	PWM A, PWM B, PWM C, PWM D, M1,	2			V
V <sub>IL</sub>	Low-level input voltage	M2, M3, RESET_AB, RESET_CD			0.8	V
I <sub>Leakage</sub>	Input leakage current				100	μA
OTW/SHUTD	OWN (SD)					
R <sub>INT_PU</sub>	Internal pullup resistance, OTW to VREG, SD to VREG		20	26	32	kΩ
M		Internal pullup resistor 3 3.3	3.3	3.6	V	
V <sub>OH</sub>	High-level output voltage	External pullup of 4.7 k $\Omega$ to 5 V	4.5		5	v
V <sub>OL</sub>	Low-level output voltage	I <sub>O</sub> = 4 mA		0.2	0.4	V
FANOUT	Device fanout OTW, SD	No external pullup		30		Devices

## TYPICAL CHARACTERISTICS, BTL CONFIGURATION



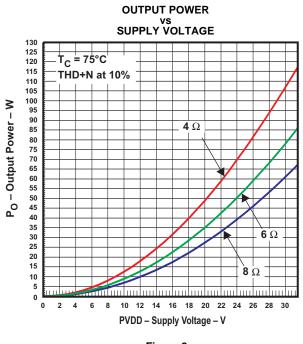
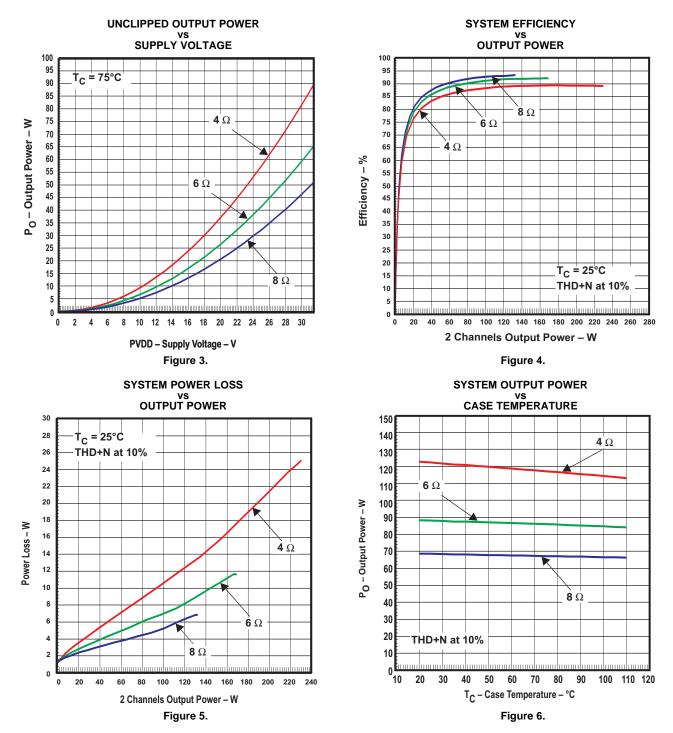


Figure 2.

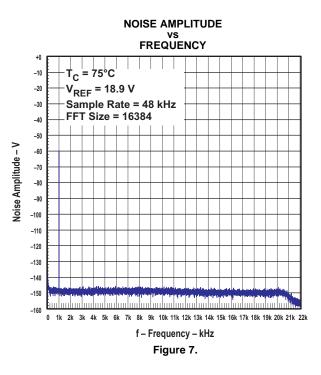




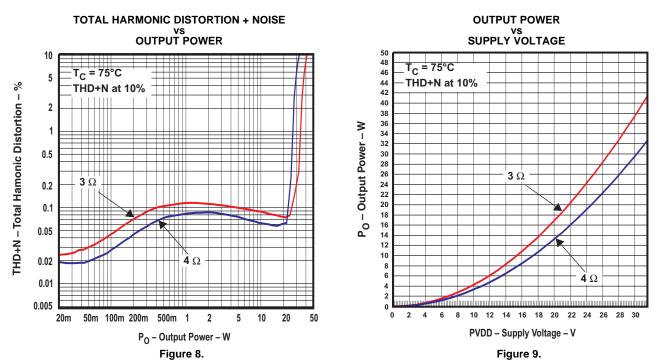




## **TYPICAL CHARACTERISTICS, BTL CONFIGURATION (continued)**

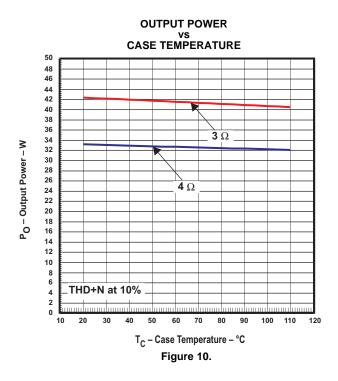




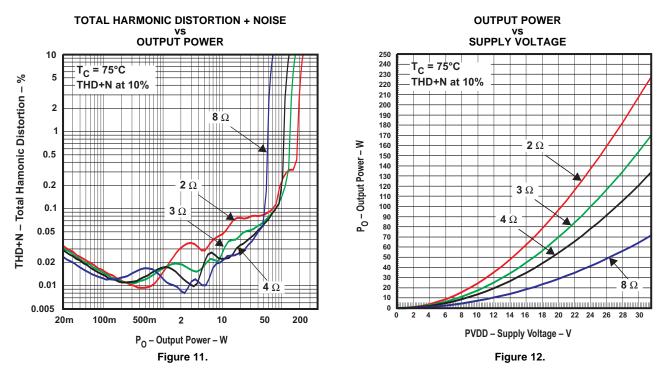




## **TYPICAL CHARACTERISTICS, SE CONFIGURATION (continued)**

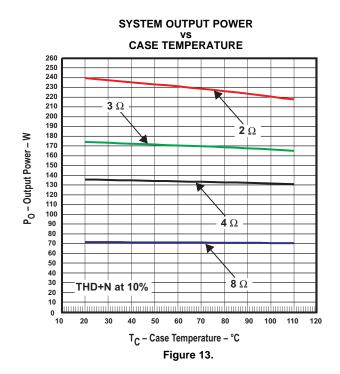


## **TYPICAL CHARACTERISTICS, PBTL CONFIGURATION**





## **TYPICAL CHARACTERISTICS, PBTL CONFIGURATION (continued)**





## **APPLICATION INFORMATION**

#### PCB Material Recommendation

FR-4 Glass Epoxy material with 2 oz. (70 µm) is recommended for use with the TAS5342. The use of this material can provide for higher power output, improved thermal performance, and better EMI margin (due to lower PCB trace inductance.

#### **PVDD Capacitor Recommendation**

The large capacitors used in conjunction with each full-birdge, are referred to as the PVDD Capacitors. These capacitors should be selected for proper voltage margin and adequate capacitance to support the power requirements. In practice, with a well designed system power supply, 1000  $\mu$ F, 50-V will support more applications. The PVDD capacitors should be low ESR type because they are used in a circuit associtated with high-speed switching.

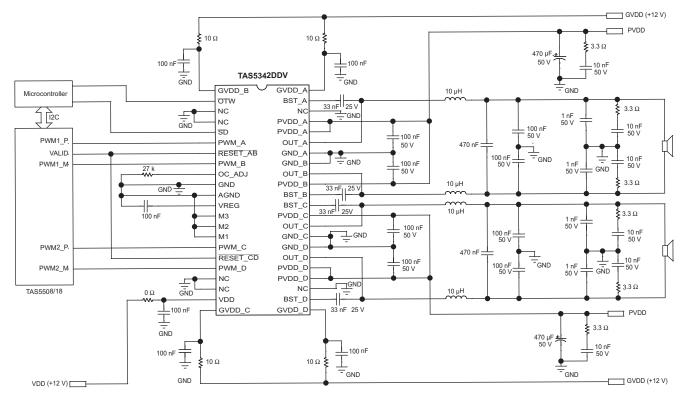
#### **Decoupling Capacitor Recommendations**

In order to design an amplifier that has robust performance, passes regulatory requirements, and exhibits good audio performance, good quality decoupling capacitors should be used. In practice, X7R should be used in this application.

The voltage of the decoupling capactors should be selected in accordance with good design practices. Temperature, ripple current, and voltage overshoot must be considered. This fact is particularly true in the selection of the  $0.1\mu$ F that is placed on the power supply to each half-bridge. It must withstand the voltage overshoot of the PWM switching, the heat generated by the amplifier during high power output, and the ripple current created by high power output. A minimum voltage rating of **50-V** is required for use with a 31.5-V power supply.

#### **System Design Recommendations**

The following schematics and PCB layouts illustrate "best practices" in the use of the TAS5342.







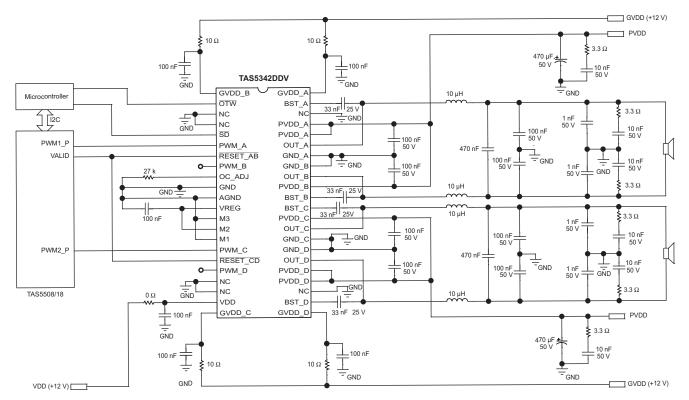
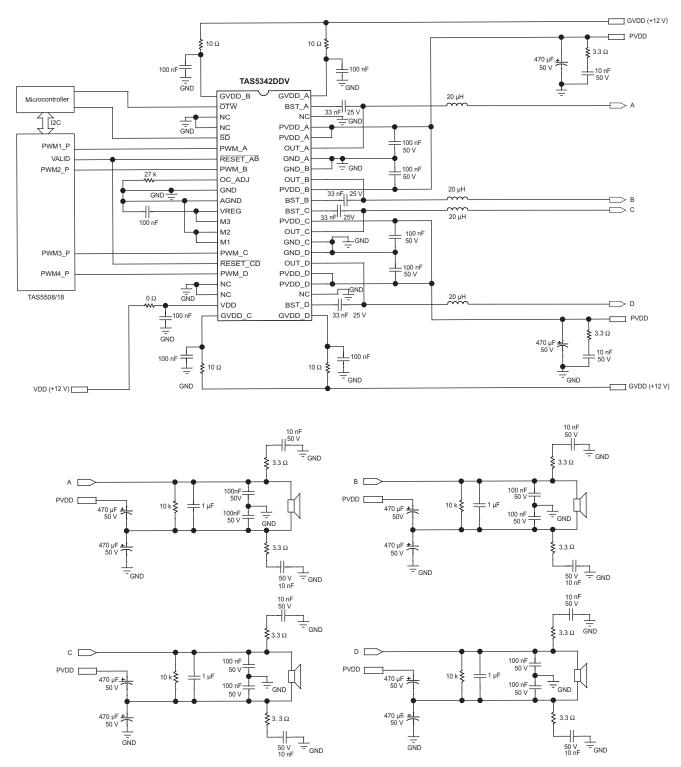


Figure 15. Typical Non-Differential (1N) BTL Application With AD Modulation Filters









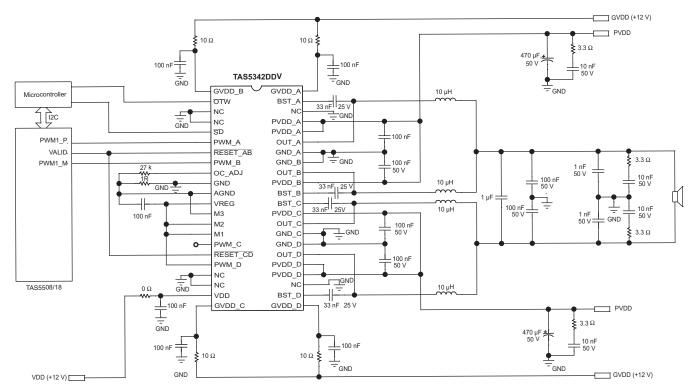
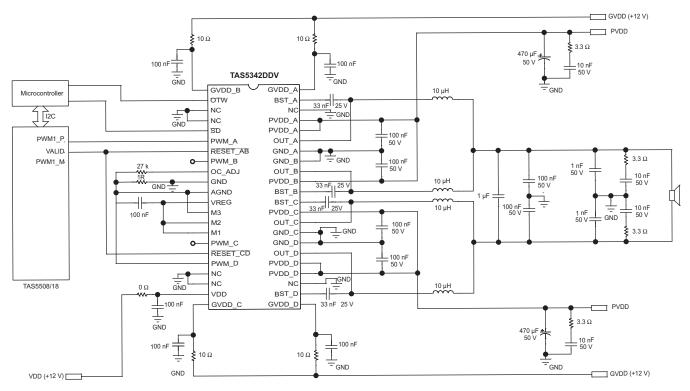
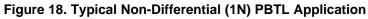


Figure 17. Typical Differential (2N) PBTL Application With AD Modulation Filters







## THEORY OF OPERATION

#### **POWER SUPPLIES**

To facilitate system design, the TAS5342 needs only a 12-V supply in addition to the (typical) 31.5-V power-stage supply. An internal voltage regulator provides suitable voltage levels for the digital and low-voltage analog circuitry. Additionally, all circuitry requiring a floating voltage supply, e.g., the high-side gate drive, is accommodated by built-in bootstrap circuitry requiring only an external capacitor for each half-bridge.

In order to provide outstanding electrical and acoustical characteristics, the PWM signal path including gate drive and output stage is designed as identical, independent half-bridges. For this reason, each half-bridge has separate gate drive supply (GVDD X), bootstrap pins (BST X), and power-stage supply pins (PVDD X). Furthermore, an additional pin (VDD) is provided as supply for all common circuits. Although supplied from the same 12-V source, it is highly recommended to separate GVDD A, GVDD B, GVDD C, GVDD D, and VDD on the printed-circuit board (PCB) by RC filters (see application diagram for details). These RC filters provide the recommended high-frequency isolation. Special attention should be paid to placing all decoupling capacitors as close to their associated pins as possible. In general, inductance between the power supply pins and decoupling capacitors must be avoided. (See reference board documentation for additional information.)

For a properly functioning bootstrap circuit, a small ceramic capacitor must be connected from each bootstrap pin (BST X) to the power-stage output pin (OUT X). When the power-stage output is low, the bootstrap capacitor is charged through an internal diode connected between the gate-drive powersupply pin (GVDD\_X) and the bootstrap pin. When the power-stage output is high, the bootstrap capacitor potential is shifted above the output potential and thus provides a suitable voltage supply for the high-side gate driver. In an application with PWM switching frequencies in the range from 352 kHz to 384 kHz, it is recommended to use 33-nF ceramic capacitors, size 0603 or 0805, for the bootstrap supply. These 33-nF capacitors ensure sufficient energy storage, even during minimal PWM duty cycles, to keep the high-side power stage FET (LDMOS) fully turned on during the remaining part of the PWM cycle. In an application running at a reduced switching frequency, generally 192 kHz, the bootstrap capacitor might need to be increased in value.

Special attention should be paid to the power-stage power supply; this includes component selection, PCB placement, and routing. As indicated, each half-bridge has independent power-stage supply pins (PVDD\_X). For optimal electrical performance, EMI compliance, and system reliability, it is important that each PVDD\_X pin is decoupled with a 100-nF ceramic capacitor placed as close as possible to each supply pin. It is recommended to follow the PCB layout of the TAS5342 reference design. For additional information on recommended power supply and required components, see the application diagrams given previously in this data sheet.

The 12-V supply should be from a low-noise, low-output-impedance voltage regulator. Likewise, the 31.5-V power-stage supply is assumed to have low output impedance and low noise. The power-supply sequence is not critical as facilitated by the internal power-on-reset circuit. Moreover, the TAS5342 is fully protected against erroneous power-stage turnon due to parasitic gate charging. Thus, voltage-supply ramp rates (dV/dt) are non-critical within the specified range (see the *Recommended Operating Conditions* section of this data sheet).

#### SYSTEM POWER-UP/POWER-DOWN SEQUENCE

#### Powering Up

The TAS5342 does not require a power-up sequence. The outputs of the H-bridges remain in a high-impedance state until the gate-drive supply voltage (GVDD\_X) and VDD voltage are above the undervoltage protection (UVP) voltage threshold (see the *Electrical Characteristics* section of this data sheet). Although not specifically required, it is recommended to hold RESET\_AB and RESET\_CD in a low state while powering up the device. This allows an internal circuit to charge the external bootstrap capacitors by enabling a weak pulldown of the half-bridge output.

When the TAS5342 is being used with TI PWM modulators such as the TAS5518, no special attention to the state of RESET\_AB and RESET\_CD is required, provided that the chipset is configured as recommended.

#### **Powering Down**

The TAS5342 does not require a power-down sequence. The device remains fully operational as long as the gate-drive supply (GVDD\_X) voltage and VDD voltage are above the undervoltage protection (UVP) voltage threshold (see the *Electrical* 



*Characteristics* section of this data sheet). Although not specifically required, it is a good practice to hold RESET\_AB and RESET\_CD low during power down, thus preventing audible artifacts including pops or clicks.

When the TAS5342 is being used with TI PWM modulators such as the TAS5518, no special attention to the state of RESET\_AB and RESET\_CD is required, provided that the chipset is configured as recommended.

## Mid Z Sequence Compatability

The TAS5342 is compatable with the Mid Z sequence of the TAS5086 Modulator. The Mid Z Sequence is a series of pulses that is generated by the modulator. This sequence causes the power stage to slowly enable its outputs as it begins to switch.

By slowly starting the PWM switching, the impulse response created by the onset of switching is reduced. This impulse response is the acoustic artifact that is heard in the output transducers (loudspeakers) and is commonly termed "click" or "pop".

The low acoustic artifact noise of the TAS5342 will be further decreased when used in conjunction with the TAS5086 modulator with the Mid Z Sequence enabled.

The Mid Z sequence is primarily used for the single-ended output configuration. It facilitates a "softer" PWM output start after the split cap output configuration is charged.

## ERROR REPORTING

The SD and OTW pins are both active-low, open-drain outputs. Their function is for protection-mode signaling to a PWM controller or other system-control device.

Any fault resulting in device shutdown is signaled by the  $\overline{SD}$  pin going low. Likewise,  $\overline{OTW}$  goes low when the device junction temperature exceeds 125°C (see the following table).

SD	στω	DESCRIPTION
0	0	Overtemperature (OTE) or overload (OLP) or undervoltage (UVP)
0	1	Overload (OLP) or undervoltage (UVP)
1	0	Junction temperature higher than 125°C (overtemperature warning)
1	1	Junction temperature lower than 125°C and no OLP or UVP faults (normal operation)

Note that asserting either RESET\_AB or RESET\_CD low forces the SD signal high, independent of faults being present. TI recommends monitoring the OTW signal using the system microcontroller and responding to an overtemperature warning signal by, e.g., turning down the volume to prevent further heating of the device resulting in device shutdown (OTE).

To reduce external component count, an internal pullup resistor to 3.3 V is provided on both SD and OTW outputs. Level compliance for 5-V logic can be obtained by adding external pullup resistors to 5 V (see the *Electrical Characteristics* section of this data sheet for further specifications).

#### **DEVICE PROTECTION SYSTEM**

The TAS5342 contains advanced protection circuitry carefully designed to facilitate system integration and ease of use, as well as to safeguard the device from permanent failure due to a wide range of fault conditions such as short circuits, overload, overtemperature, and undervoltage. The TAS5342 responds to a fault by immediately setting the power stage in a high-impedance (Hi-Z) state and asserting the SD pin low. In situations other than overload and (OTE), over-temperature error the device automatically recovers when the fault condition has been removed, i.e., the supply voltage has increased.

The device will function on errors, as shown in the following table.

BTL	MODE	PBT	L MODE	SE MODE			
Local Error In	Turns Off	Local Error In	Turns Off	Local Error In	Turns Off		
Α	A + B	А		А	A + B		
В	Атб	В	A + B + C	В	ATD		
С	C + D	С	+ D	С	C + D		
D	C+D	D		D	C + D		

Bootstrap UVP does not shutdown according to the table, it shutsdown the respective halfbridge.

# Use of TAS5342 in High-Modulation-Index Capable Systems

This device requires at least 30 ns of low time on the output per 384-kHz PWM frame rate in order to keep the bootstrap capacitors charged. As an example, if the modulation index is set to 99.2% in the TAS5508, this setting allows PWM pulse durations down to 10 ns. This signal, which does not meet the 30-ns requirement, is sent to the PWM\_X pin and this low-state pulse time does not allow the bootstrap capacitor to stay charged. The TAS5342 device requires limiting the TAS5508 modulation index to 97.7% to keep the bootstrap capacitor charged under all signals and loads.

The TAS5342 contains a bootstrap capacitor under voltage protection circuit (BST\_UVP) that monitors the voltage on the bootstrap capacitors. When the



voltage on the bootstrap capacitors is less than required for proper control of the High-Side MOSFETs, the device will initiate bootstrap capacitor recharge sequences until the bootstrap capacitors are properly charged for robust operation. This function may be activated with PWM pulses less than 30 nS.

Therefore, TI strongly recommends using a TI PWM processor, such as TAS5518, TAS5086 or TAS5508, with the modulation index set at 97.7% to interface with TAS5342.

# Overcurrent (OC) Protection With Current Limiting and Overload Detection

The device has independent, fast-reacting current detectors with programmable trip threshold (OC threshold) on all high-side and low-side power-stage FETs. See the following table for OC-adjust resistor values. The detector outputs are closely monitored by two protection systems. The first protection system controls the power stage in order to prevent the output current from further increasing, i.e., it performs a current-limiting function rather than prematurely shutting down during combinations of high-level music transients and extreme speaker load impedance drops. If the high-current situation persists, i.e., the power stage is being overloaded, a second protection system triggers a latching shutdown, resulting in the power stage being set in the high-impedance (Hi-Z) state. Current limiting and overload protection are independent for half-bridges A and B and, respectively, C and D. That is, if the bridge-tied load between half-bridges A and B causes an overload fault, only half-bridges A and B are shut down.

- For the lowest-cost bill of materials in terms of component selection, the OC threshold measure should be limited, considering the power output requirement and minimum load impedance. Higher-impedance loads require a lower OC threshold.
- The demodulation-filter inductor must retain at least 5 µH of inductance at twice the OC threshold setting.

Unfortunately, most inductors have decreasing inductance with increasing temperature and increasing current (saturation). To some degree, an increase in temperature naturally occurs when operating at high output currents, due to core losses and the dc resistance of the inductor's copper winding. A thorough analysis of inductor saturation and thermal properties is strongly recommended.

Setting the OC threshold too low might cause issues such as lack of enough output power and/or unexpected shutdowns due to too-sensitive overload detection. In general, it is recommended to follow closely the external component selection and PCB layout as given in the *Application* section.

For added flexibility, the OC threshold is programmable within a limited range using a single external resistor connected between the OC ADJ pin and AGND. (See the Electrical Characteristics section of this data sheet for information on the correlation between programming-resistor value and the OC threshold.) It should be noted that a properly functioning overcurrent detector assumes the presence of a properly designed demodulation filter at the power-stage output. Short-circuit protection is not provided directly at the output pins of the power stage but only on the speaker terminals (after the demodulation filter). It is required to follow certain guidelines when selecting the OC threshold and an appropriate demodulation inductor:

OC-Adjust Resistor Values (kΩ)	Max. Current Before OC Occurs (A), T <sub>C</sub> =75°C
27	10.1
33	9.1
47	7.1

The reported max peak current in the table above is measured with continuous current in 1  $\Omega$ , one channel active and the other one muted.

# Pin-To-Pin Short Circuit Protection System (PPSC)

The PPSC detection system protects the device from permanent damage in the case that a power output pin (OUT X) is shorted to GND X or PVDD X. For comparison the OC protection system detects an over current after the demodulation filter where PPSC detects shorts directly at the pin before the filter. PPSC detection is performed at startup i.e. when VDD is supplied, consequently a short to either GND\_X or PVDD\_X after system startup will not activate the PPSC detection system. When PPSC detection is activated by a short on the output, all half bridges are kept in a Hi-Z state until the short is removed, the device then continues the startup sequence and starts switching. The detection is controlled globally by a two step sequence. The first step ensures that there are no shorts from OUT X to GND\_X, the second step tests that there are no shorts from OUT X to PVDD X. The total duration of this process is roughly proportional to the capacitance of the output LC filter. The typical duration is < 15 ms/µF. While the PPSC detection is in progress, SD is kept low, and the device will not react to changes applied to the RESET pins. If no shorts are present the PPSC detection passes, and SD is released. A device reset will not start a new PPSC detection.



PPSC detection is enabled in BTL and PBTL output configurations, the detection is not performed in SE mode. To make sure not to trip the PPSC detection system it is recommended not to insert resistive load to GND X or PVDD X.

#### **Overtemperature Protection**

The TAS5342 has a two-level temperature-protection system that asserts an active-low warning signal (OTW) when the device junction temperature exceeds 125°C (nominal) and, if the device junction temperature exceeds 155°C (nominal), the device is put into thermal shutdown, resulting in all half-bridge outputs being set in the high-impedance (Hi-Z) state and SD being asserted low. OTE is latched in this case. To clear the OTE latch, either RESET\_AB or RESET\_CD must be asserted. Thereafter, the device resumes normal operation.

# Undervoltage Protection (UVP) and Power-On Reset (POR)

The UVP and POR circuits of the TAS5342 fully protect the device in any power-up/down and brownout situation. While powering up, the POR circuit resets the overload circuit (OLP) and ensures that all circuits are fully operational when the GVDD\_X and VDD supply voltages reach stated in the Electrical Characteristics Table. Although GVDD\_X and VDD are independently monitored, a supply voltage drop below the UVP threshold on any VDD or GVDD\_X pin results in all half-bridge outputs immediately being set in the high-impedance (Hi-Z) state and SD being asserted low. The device automatically resumes operation when all supply voltages have increased above the UVP threshold.

#### DEVICE RESET

Two reset pins are provided for independent control of half-bridges A/B and C/D. When RESET\_AB is asserted low, all four power-stage FETs in halfbridges A and B are forced into a high-impedance (Hi-Z) state. Likewise, asserting RESET\_CD low forces all four power-stage FETs in half-bridges C and D into a high-impedance state. Thus, both reset pins are well suited for hard-muting the power stage if needed.

In BTL modes, to accommodate bootstrap charging prior to switching start, asserting the reset inputs low enables weak pulldown of the half-bridge outputs. In the SE mode, the weak pulldowns are not enabled, and it is therefore recommended to ensure bootstrap capacitor charging by providing a low pulse on the PWM inputs when reset is asserted high.

Asserting either reset input low removes any fault information to be signalled on the SD output, i.e., SD is forced high.

A rising-edge transition on either reset input allows the device to resume operation after an overload fault. To ensure thermal reliability, the rising edge of reset must occur no sooner than 4ms after the falling edge of SD.



### **PACKAGING INFORMATION**

Orderable Device	Status <sup>(1)</sup>	Package Type	Package Drawing	Pins	Package Qty	Eco Plan <sup>(2)</sup>	Lead/ Ball Finish	MSL Peak Temp <sup>(3)</sup>	Samples (Requires Login)
TAS5342DDV	NRND	HTSSOP	DDV	44	35	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	
TAS5342DDVG4	NRND	HTSSOP	DDV	44	35	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	
TAS5342DDVR	NRND	HTSSOP	DDV	44	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	
TAS5342DDVRG4	NRND	HTSSOP	DDV	44	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	

<sup>(1)</sup> The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

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**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

<sup>(3)</sup> MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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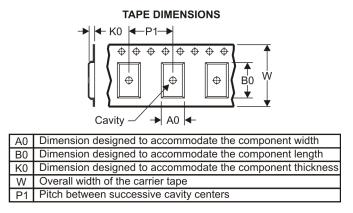
## PACKAGE MATERIALS INFORMATION

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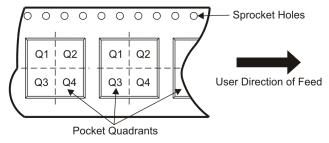
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## TAPE AND REEL INFORMATION





### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



Device	Package Type	Package Drawing	Pins		Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TAS5342DDVR	HTSSOP	DDV	44	2000	330.0	24.4	8.6	15.6	1.8	12.0	24.0	Q1

TEXAS INSTRUMENTS

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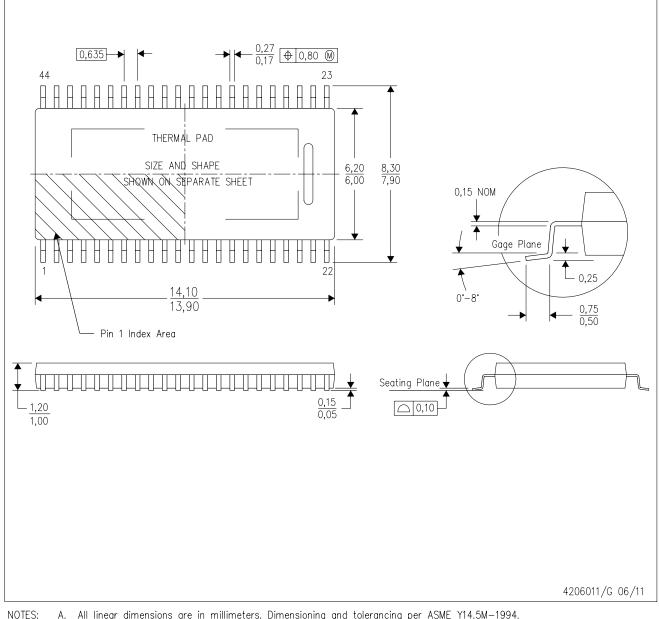
25-May-2009



\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TAS5342DDVR	HTSSOP	DDV	44	2000	346.0	346.0	41.0

DDV (R-PDSO-G44) PowerPAD<sup>™</sup> PLASTIC SMALL OUTLINE PACKAGE (DIE DOWN)



A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.
B. This drawing is subject to change without notice.

- c. Body dimensions do not include mold flash or protrusion not to exceed 0,15.
- D. This package is designed to be attached directly to an external heatsink. Refer to Technical Brief, PowerPad Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 for information regarding recommended board layout. This document is available at www.ti.com <a href="https://www.ti.com">http://www.ti.com</a>. See the product data sheet for details regarding the exposed thermal pad dimensions.

PowerPAD is a trademark of Texas Instruments.



# ™SMALL OUTLINE <u>PACKAGE</u> PowerPAD DDV (R-PDSO-G44) THERMAL INFORMATION This PowerPAD<sup>™</sup> package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. This design optimizes the heat transfer from the integrated circuit (IC). For additional information on the PowerPAD package and how to take advantage of its heat dissipating abilities, refer to Technical Brief, PowerPAD Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 and Application Brief, PowerPAD Made Easy, Texas Instruments Literature No. SLMA004. Both documents are available at www.ti.com. The exposed thermal pad dimensions for this package are shown in the following illustration. 44 23 Exposed Thermal 38 Pad 4,20 3,50 2x0,60 22 1 8,30 6.90 Top View Exposed Thermal Pad Dimensions 4206975-2/D 07/11

NOTE: All linear dimensions are in millimeters



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DSP	dsp.ti.com	Energy and Lighting	www.ti.com/energy
Clocks and Timers	www.ti.com/clocks	Industrial	www.ti.com/industrial
Interface	interface.ti.com	Medical	www.ti.com/medical
Logic	logic.ti.com	Security	www.ti.com/security
Power Mgmt	power.ti.com	Space, Avionics and Defense	www.ti.com/space-avionics-defense
Microcontrollers	microcontroller.ti.com	Video and Imaging	www.ti.com/video
RFID	www.ti-rfid.com		
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