



June 1997-3

#### **FEATURES**

- Meets CCITT G.703 Pulse Mask Template for 2.048Mbps (E1) Rates
- Transmitter and Receiver Interfaces Can Be:
  - Single Ended,  $75\Omega$  Capacitive or Transformer Coupled
  - Balanced, 100 or 120Ω Transformer Coupled
- Minimum Return Loss is 20dB (Receive) and 18dB (Transmit), Exceeds G.703 and ETSI 300 166 Specifications
- Bipolar Outputs can be Disabled Individually (High Z Outputs)
- System Interface is TTL Compatible on Digital Input and TTL/CMOS Compatible On Digital Output Pins

- Individual Channel Loss of Signal Detection, Local and Remote Digital Loopback
- Fifth Driver For Monitoring and Testing
- Low Power, CMOS Technology
- Over-temperature Protection

#### **APPLICATIONS**

- Multi-Line E1 Interface Cards
- E1 Network Equipment
  - Multiplexers
  - Cross Connects
  - Switching Systems
- Fault Tolerant Systems

### **GENERAL DESCRIPTION**

The XR-T5794 is an optimized line interface unit, built using low power CMOS technology. The device contains four independent E1 channels for primary rate, PCM applications up to 2.048Mbps. Each channel performs the driver and receiver functions necessary to convert bipolar signals to TTL/CMOS compatible logic levels and vice versa. The device supports single ended or balanced line interfaces on each channel, thereby providing the user an option of reducing system cost and board space by replacing the transformer with a capacitor.

Each of the four drivers can be independently disabled, allowing maximum flexibility in system power management. Output pulses are fully CCITT G.703 compliant. Moreover, the return loss is at least 18dB over a frequency range of 51kHz to 3.072MHz.

The slicing circuit in the receive path is able to tolerate a maximum of 12dB of cable loss with a minimum input

sensitivity of 600mV over the operating temperature range. Return loss on the receive interfaces is minimum 20dB from 51kHz to 3.072MHz.

Local and remote loopbacks can be performed on any of the four channels. A separate loss of signal (LOS) detection circuitry and a LOS pin is provided for each input. A fifth transmitter has been provided to support dedicated monitoring and testing purposes on any of the eight bipolar paths. For designers not requiring the fifth (monitor) driver, EXAR offers the XR-T5793, a pin compatible version of the XR-T5794.

The XR-T5794 is targeted for multi-line E1 line card applications where real estate, low power consumption and back-up redundancy are critical. Also, the device may be used in T1 applications (1.544Mbps) which do not require meeting the DSX-1 cross connect pulse template.

#### ORDERING INFORMATION

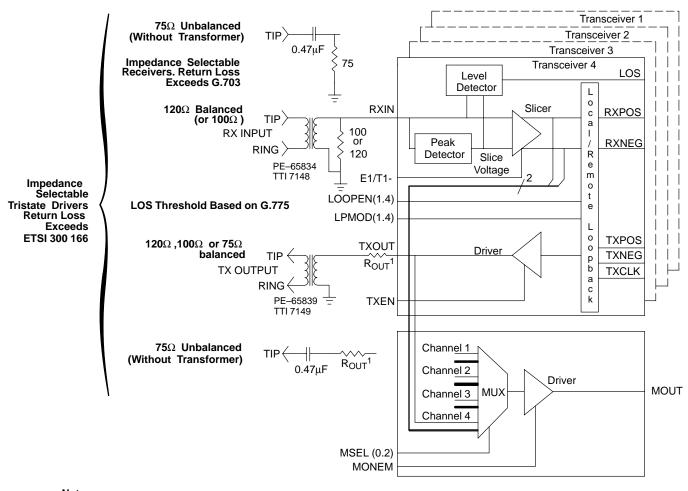
Part No.	Package	Operating Temperature Range		
XR-T5794IJ	68 Lead PLCC	-40°C to +85°C		
XR-T5794IV	80 Lead TQFP (14x14x1.4 mm)	-40°C to +85°C		

Rev. 2.02

T**⊙**M<sup>™</sup>



### **BLOCK DIAGRAM**

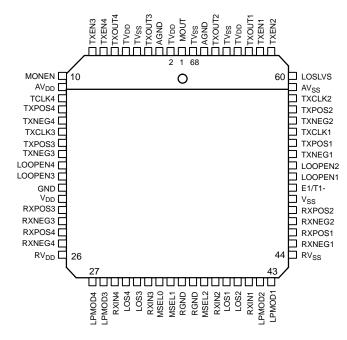


**Note**  $^{1}R_{OUT} = 68\Omega$  for  $120\Omega$  line impedance,  $R_{OUT} = 62\Omega$  for  $100\Omega$  line impedance,  $R_{OUT} = 68\Omega$  for  $75\Omega$  line impedance

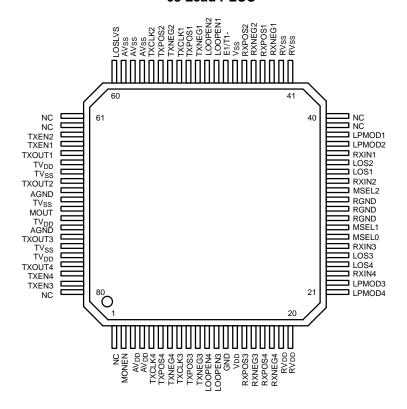
Figure 1. Block Diagram



#### **PIN CONFIGURATION**



# 68 Lead PLCC



80 Lead TQFP (14 x 14 x 1.4 mm)



# **PIN DESCRIPTION**

PLCC Pin #	SQFP Pin #	Symbol	Туре	Description		
1	71	MOUT	0	Signal Monitor Output. If MONEN=1, this output tracks the selected signal. Hi-Z otherwise. The channel selection is done using MONSEL[20] inputs.		
2	72	$TV_DD$	$V_{DD}$	Transmit $V_{DD.}$ + 5V (±5%).		
3	73	AGND	GND	Analog Ground.		
4	74	TXOUT3	0	Transmitter 3 Output. Transmitter 3 bipolar output connected to coupling capacitor or pulse transformer by a resistor.		
5	75	$TV_SS$	$V_{SS}$	Transmit $V_{SS.}$ -5V (±5%).		
6	76	$TV_DD$	$V_{DD}$	Transmit $V_{DD.}$ +5V (±5%).		
7	77	TXOUT4	0	<b>Transmitter 4 Output.</b> Transmitter 4 bipolar output connected to coupling capacitor or pulse transformer by a resistor.		
8	78	TXEN4	I	<b>Transmitter 4 Output Enable.</b> If driven high the transmitter 4 output drivers are enabled. Hi-Z otherwise.		
9	79	TXEN3	I	<b>Transmitter 3 Output Enable.</b> If driven high the transmitter 3 output drivers are enabled. Hi-Z otherwise.		
10	2	MONEN	I	<b>Monitor/Test Output Enable.</b> If driven high the output driver of the MOUT output is enabled. Hi-Z otherwise.		
11	3,4	$AV_DD$	$V_{DD}$	Analog V <sub>DD.</sub>		
12	5	TXCLK4	I	Transmitter 4 Clock Input. Apply logic one when RZ signals are supplied to data inputs.		
13	6	TXPOS4	I	<b>Transmitter 4 Positive Data In.</b> Positive data input in NRZ or RZ format for transmitter 4.		
14	7	TXNEG4	I	<b>Transmitter 4 Negative Data In.</b> Negative data input in NRZ or RZ format for transmitter 4.		
15	8	TXCLK3	I	Transmitter 3 Clock Input. Apply logic one when RZ signals are supplied to data inputs.		
16	9	TXPOS3	I	<b>Transmitter 3 Positive Data In.</b> Positive data input in NRZ or RZ format for transmitter 3.		
17	10	TXNEG3	I	<b>Transmitter 3 Negative Data In.</b> Negative data input in NRZ or RZ format for transmitter 3.		
18	11	LOOPEN4	I	<b>Loop Enable 4.</b> If driven high the specified loop type will be enabled for channel 4. Otherwise normal operation will continue.		
19	12	LOOPEN3	I	<b>Loop Enable 3.</b> If driven high the specified loop type will be enabled for channel 3. Otherwise normal operation will continue.		
20	13	GND	GND	Digital Ground.		
21	14	$V_{DD}$	$V_{DD}$	<b>Digital V<sub>DD.</sub></b> +5V (±5%).		
22	15	RXPOS3	0	Receiver 3 Positive Data Out. Positive data output in NRZ or RZ format for receiver 3.		
23	16	RXNEG3	0	<b>Receiver 3 Negative Data Out.</b> Negative data output in NRZ or RZ format for receiver 3.		
24	17	RXPOS4	0	Receiver 4 Positive Data Out. Positive data output in NRZ or RZ format for receiver 4.		



# PIN DESCRIPTION (CONT'D)

PLCC Pin #	SQFP Pin #	Symbol	Туре	Description			
25	18	RXNEG4	0	<b>Receiver 4 Negative Data Out.</b> Negative data output in NRZ or RZ format for receiver 4.			
26	19, 20	$RV_{DD}$	$V_{DD}$	Receive $V_{DD.}$ +5V (±5%).			
27	21	LPMOD4	I	<b>Loop Mode 4.</b> If driven high the loopback mode of channel 4 will be set to remote loop. Otherwise theloopback mode will remain at local loop. The actual loopback will be activated when the LOOPEN4 is asserted.			
28	22	LPMOD3	I	<b>Loop Mode 3.</b> If driven high the loopback mode of channel 3 will be set to remote loop. Otherwise the loopback mode will remain at local loop. The actual loopback will be activated when the LOOPEN3 is asserted.			
29	23	RXIN4	I	<b>Receiver 4 Input.</b> Receiver 4 bipolar input connected to coupling capacitor or pulse transformer.			
30	24	LOS4	0	Receiver 4 Loss Of Signal. Asserted during LOS condition. Clear otherwise.			
31	25	LOS3	0	Receiver 3 Loss Of Signal. Asserted during LOS condition. Clear otherwise.			
32	26	RXIN3	I	<b>Receiver 3 Input.</b> Receiver 3 bipolar input connected to coupling capacitor or pulse transformer.			
33	27	MSEL0	I	<b>Monitor Channel Select 0.</b> Select line, used to select a channel for monitoring using the MOUT pin based on the following assignment:			
				MSEL2 MSEL1 MSEL0 SELECTS			
				0 0 Line 1 Receive			
				0 0 1 Line 2 Receive 0 1 0 Line 3 Receive			
				0 1 1 Line 4 Receive			
				1 0 0 Line 1 Transmit			
				1 0 1 Line 2 Transmit 1 1 0 Line 3 Transmit			
				1 1 1 Line 4 Transmit			
				<b>Note</b> The monitoring is only done on the NRZ data output signals from the receiver or from the transmitter line side.			
34	28	MSEL1	I	Monitor Channel Select 1. See table above.			
35	29, 30	RGND	GND	Receive Ground.			
36	31	RGND	GND	Receive Ground.			
37	32	MSEL2	I	Monitor Channel Select 2. See table above.			
38	33	RXIN2	I	<b>Receiver 2 Input</b> . Receiver 2 bipolar input connected to coupling capacitor or pulse transformer.			
39	34	LOS1	0	Receiver 1 Loss Of Signal. Asserted during LOS condition. Clear otherwise.			
40	35	LOS2	0	Receiver 2 Loss Of Signal. Asserted during LOS condition. Clear otherwise.			
41	36	RXIN1	I	<b>Receiver 1 Input</b> . Receiver 1 bipolar input connected to coupling capacitor or pulse transformer.			
42	37	LPMOD2	-	<b>Loop Mode 2.</b> If driven high the loopback mode of channel 2 will be set to remote loop. Otherwise the loopback mode will remain at local loop. The actual loopback will be activated when the LOOPEN2 is asserted.			



# PIN DESCRIPTION (CONT'D)

PLCC Pin #	SQFP Pin #	Symbol	Туре	Description	
43	38	LPMOD1	I	<b>Loop Mode 1.</b> lif driven high the loopback mode of channel 1 will be set to remote loop. Otherwise the loopback mode will remain at local loop. The actual loopback will be activated when the LOOPEN1 is asserted.	
44	41, 42	RV <sub>SS</sub>	$V_{SS}$	<b>Receive V<sub>SS</sub>.</b> -5V (±5%).	
45	43	RXNEG1	0	Receiver 1 Negative Data Out. Negative data output in NRZ or RZ format for receiver 1.	
46	44	RXPOS1	0	Receiver 1 Positive Data Out. Positive data output in NRZ or RZ format for receiver 1.	
47	45	RXNEG2	0	Receiver 2 Negative Data Out. Negative data output in NRZ or RZ format for receiver 2.	
48	46	RXPOS2	0	Receiver 2 Positive Data Out. Positive data output in NRZ or RZ format for receiver 2.	
49	47	$V_{SS}$	$V_{SS}$	<b>Digital V<sub>SS.</sub></b> -5V (±5%).	
50	48	E1/T1-	I	<b>E1/T1- Selection.</b> Apply logic one to select the receive data threshold appropriate for E1 operation. Connect to ground to select the T1 data threshold.	
51	49	LOOPEN1	I	<b>Loop Enable 1.</b> If driven high the specified loopback mode will be enabled for channel 1. Otherwise normal operation will continue.	
52	50	LOOPEN2	I	<b>Loop Enable 2.</b> If driven high the specified loopback mode will be enabled for channel 2. Otherwise normal operation will continue.	
53	51	TXNEG1	I	<b>Transmitter 1 Negative Data In.</b> Negative data input in NRZ or RZ format fo transmitter 1.	
54	52	TXPOS1	I	<b>Transmitter 1 Positive Data In.</b> Positive data input in NRZ or RZ format for transmitter 1.	
55	53	TXCLK1	I	<b>Transmitter 1 Clock Input.</b> Apply logic one when RZ signals are supplied to data inputs.	
56	54	TXNEG2	I	<b>Transmitter 2 Negative Data In.</b> Negative data input in NRZ or RZ format for transmitter 2.	
57	55	TXPOS2	I	<b>Transmitter 2 Positive Data In.</b> Positive data input in NRZ or RZ format for transmitter 2.	
58	56	TXCLK2	I	<b>Transmitter 2 Clock Input.</b> Apply logic one when RZ signals are supplied to data inputs.	
59	57, 58, 59	AV <sub>SS</sub>	$V_{SS}$	Analog V <sub>SS</sub> .	
60	60	LOSLVS	I	Loss of Signal Voltage Select. Apply logic one to select LOS voltage level appropriate for $120\Omega$ balanced receiver operation. Connect to ground to choose LOS voltage for $75\Omega$ unbalanced operation.	
61	63	TXEN2	I	<b>Transmitter 2 Output Enable.</b> If asserted the transmitter 2 output drivers are enabled. High-Z otherwise.	
62	64	TXEN1	I	<b>Transmitter 1 Output Enable.</b> If asserted the transmitter 1 output drivers are enabled. High-Z otherwise.	
63	65	TXOUT1	0	<b>Transmitter 1 Output.</b> Transmitter 1 bipolar output connected to coupling capacitor or pulse transformer through a resistor.	
64	66	$TV_DD$	$V_{DD}$	Transmit $V_{DD.}$ +5V (±5%).	
65	67	TV <sub>SS</sub>	$V_{SS}$	Transmit $V_{SS.}$ -5 $\vee$ (±5%).	



# PIN DESCRIPTION (CONT'D)

PLCC Pin #	SQFP Pin #	Symbol	Туре	Description
66	68	TXOUT2	0	<b>Transmitter 2 Output.</b> Transmitter 2 bipolar output connected to coupling capacitor or pulse transformer through a resistor.
67	69	AGND	GND	Analog Ground.
68	70	$TV_SS$	$V_{SS}$	Transmit $V_{SS.}$ -5V (±5%).
-	1, 39, 40, 61, 62, 80	NC		No Connect.



# DC ELECTRICAL CHARACTERISTICS

**Test Conditions:**  $T_A = -40$ °C to 25°C to 85°C, all  $V_{DD}s = 5V \pm 5\%$ , all  $V_{SS}s = -5V \pm 5\%$ , all GNDs = 0V

Symbol	Parameter	Min.	Тур.	Max.	Unit	Conditions
DC Paramete	ers					
V <sub>DD</sub> s	DC Supply Positive	4.75	5.00	5.25	V	
V <sub>SS</sub> s	DC Supply Negative	-4.75	-5.00	-5.25	V	
Inputs						
V <sub>IH</sub>	High Level Input	2.0			V	
$V_{IL}$	Low Level Input			0.8	V	
I <sub>PDC</sub>	Input Pull Down Current			40	μΑ	
Outputs		_				
V <sub>OH</sub>	High Level Output	3.5			V	I <sub>OH</sub> = -10μA
V <sub>OH</sub>	High Level Output	2.4			V	$I_{OH} = -40\mu A$
$V_{OL}$	Low Level Output			0.4	V	I <sub>OL</sub> = 1.6mA
Receiver Spe	ecifications					
R <sub>XP</sub>	Receiver Sensitivity	0.6		4.2	Vp	
R <sub>XCL</sub>	Allowed Cable Loss (0dB=2.4V)	0	10 10	12 12	dB dB	1.024MHz (E1) 772kHz (T1)
R <sub>XIWT</sub>	Interference Margin (E1)	16			dB	With 6dB Cable Loss
R <sub>XTI</sub>	Receiver Slicing Level (T1) <sup>1</sup>	60	65	70	%	Peak Voltage %
R <sub>XEI</sub>	Receiver Slicing Level (E1) <sup>1</sup>	45	50	55	%	Peak Voltage %
R <sub>XLOS</sub>	Receiver LOS Threshold		0.2	0.3	V	
R <sub>IN</sub>	Input Resistance	2.5			kΩ	Up to 3.072MHz
Power Speci	fications (Without Monitor Chan	nel)				
P <sub>D</sub>	Power Dissipation		400	680	mW	
$P_{D}$	Power Dissipation		250	280	mW	All Drivers in High-Z
P <sub>C</sub>	Power Consumption $75\Omega^2$		500	833	mW	All 1's Transmit & Receive
P <sub>C</sub>	Power Consumption $100\Omega^2$		475	860	mW	All 1's Transmit & Receive
P <sub>C</sub>	Power Consumption $120\Omega^2$		450	830	mW	All 1's Transmit & Receive
PV <sub>DD</sub>	Power Supply Requirement			Pc/2 + 5mW	mW	
PV <sub>SS</sub>	Power Supply Requirement			Pc/2 - 5mW	mW	

# Notes

Bold face parameters are covered by production test and guaranteed over operating temperature range.

Specifications are subject to change without notice



<sup>&</sup>lt;sup>1</sup> Selected by E1/T1

<sup>&</sup>lt;sup>2</sup> Power consumption = power dissipation + power to the cable.



# **AC ELECTRICAL CHARACTERISTICS**

**Test Conditions:**  $T_A = -40^{\circ}\text{C}$  to  $85^{\circ}\text{C}$ , all  $V_{DD}s = 5V \pm 5\%$ , all  $V_{SS}s = -5V \pm 5\%$ , all GNDs = 0V

Symbol	Parameter	Min.	Тур.	Max.	Unit	Conditions
V <sub>TXOUT</sub>	Output Pulse Amplitude (75 $\Omega$ )	2.13	2.37	2.60	V	
$V_{TXOUT}$	Output Pulse Amplitude (120 $\Omega$ )	2.70	3.0	3.30	V	
$V_{TXOUT}$	Output Pulse Amplitude (100 $\Omega$ )	2.3	3.0	3.7	V	
$T_XPW$	Pulse Width (2.048MHz)	224	244	264	ns	Determined by TX Clock
$T_XPW$	Pulse Width (1.544MHz)	274	324	374	ns	Determined by TX Clock
PN <sub>IMB</sub>	Pos/Neg Pulse Imbalance	-5		5	%	
T <sub>1</sub>	TXCLK Clock Period (E1)		488		ns	
T <sub>2</sub>	TXCLK Clock Period (T1)		648		ns	
T <sub>3</sub>	TXCLK Duty Cycle	48	50	52	%	
T <sub>4</sub>	Data Setup Time, TDATA to TCLK	50			ns	
T <sub>5</sub>	Data Hold Time, TCLK to	50			ns	
	TDATA					
$T_{R}$	Clock Rise Time			30	ns	
T <sub>F</sub>	Clock Fall Time			30	ns	
Т <sub>6</sub>	Receive Data High (E1)	219	244	269	ns	0dB Cable Loss
T <sub>7</sub>	Data Propagation Delay			100	ns	
Т8	Receive Rise Time			50	ns	
T <sub>9</sub>	Receive Fall Time			50	ns	

#### Notes

**Bold face parameters** are covered by production test and guaranteed over operating temperature range.

# Specifications are subject to change without notice

# **ABSOLUTE MAXIMUM RATINGS**

Storage Temperature ...... -65°C to +150°C Supply Voltage .....  $\pm 7V$ 



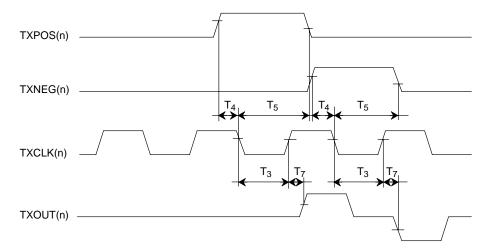


Figure 2. Transmit Timing Diagram

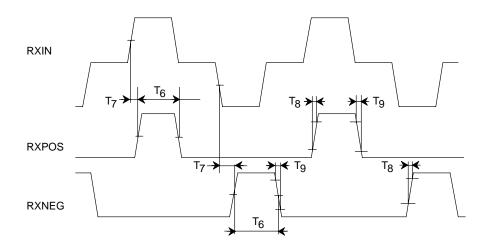


Figure 3. Receive Timing Diagram

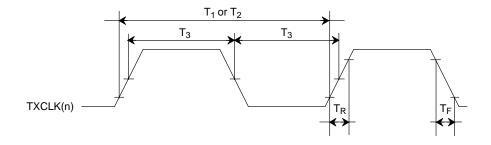


Figure 4. Transmit Clock Timing



### **RETURN LOSS REQUIREMENTS**

	75	<b>75</b> Ω		100Ω		120Ω	
Transmit Interface	Min.	Тур.	Min.	Тур.	Min.	Тур.	Units
51kHz to 102kHz	18	22	18	22	18	22	dB
102kHz to 2.048MHz	18	22	18	22	18	22	dB
2.048MHz to 3.072MHz	18	22	18	22	18	22	dB

	<b>75</b> Ω		100Ω		<b>120</b> Ω		
Receive Interface	Min.	Тур.	Min.	Тур.	Min.	Тур.	Units
51kHz to 102kHz	20	30	20	30	20	30	dB
102kHz to 2.048MHz	20	30	20	30	20	30	dB
2.048MHz to 3.072MHz	20	30	20	30	20	30	dB

#### Note

The return loss has been measured on the evaluation board coupled via a capacitor and terminated with 75 $\Omega$  impedance.

Table 1. Return Loss Requirements (Resistor Tolerance: 1% on Transmit Side, 2% on Receive Side)

Turns Ratio	Line Impedance	R <sub>LOAD</sub>
1:1	75Ω	75Ω
1:1	120Ω	120Ω
1:1	100Ω	100Ω

Turns Ratio	Line Impedance	R <sub>OUT</sub>
1:1	75Ω	68Ω
1:1.265	120Ω	68Ω
1:1.265	100Ω	62Ω

**Table 2. Input Transformer Requirements** 

**Table 3. Output Transformer Requirements** 

# **Magnetic Supplier Information:**

Pulse Telecom Product Group P.O. Box 12235 San Diego, CA 92112 Tel. (619) 674-8100 Fax. (619) 674-8262 Transpower Technologies, Inc. 24 Highway 28, Suite 202 Crystal Bay, NV 89402–0187 Tel. (702) 831–0140 Fax. (702) 831–3521





#### SYSTEM DESCRIPTION

This device is a quad E1 transceiver which provides electrical interface for 2.048Mbps applications. Its unique architecture includes four receiver circuits that convert CCITT G.703 compliant bipolar signals to TTL compatible logic levels. Likewise, in the other direction, four transmitters translate TTL compatible logic levels to G.703 compatible bipolar signals. A fifth transmitter circuit is used as a monitor output. One of the four AMI receiver inputs or transmitter outputs can be selected (via MONSEL[2..0] lines) to be monitored. The MOUT output can be disabled using the MONEN signal.

This device supports two different types of loopback functions. Each of four channels can be independently looped either in local or remote sides digitally. The remote loopback is performed between the receiver input and transmitter output. To activate the remote loopback on channel n, LOOPENn and LPMODn inputs are driven high. Local loopback on channel n, can be established similarly by driving LOOPENn high and clearing LPMODn inputs. More than one channel can be tested simultaneously.

#### **RECEIVERS**

Each of the four identical E1 line receivers will accept bipolar signals meeting the CCITT G.703 pulse mask requirements. Each input stage consists of a slicing circuitry which samples the incoming pulses at a fixed percentage of the signals maximum amplitude. The slicing voltage level is generated using a precision peak detector. The receiver section can tolerate up to 12dB of line loss (measured at 1.024MHz).

A loss of signal (LOS) is detected on any inputs by input fail circuitry. There is an independent LOS pin dedicated for each of the receivers. The LOS detection is based on signal energy instead of number of zeros.

A balanced signal ( $100\Omega$  or  $120\Omega$ ) must be coupled by a transformer. An unbalanced signal ( $75\Omega$ ) may be coupled via a capacitor or transformer coupled.

#### **TRANSMITTERS**

This device contains five identical CCITT G.703 compliant transmitters which meet the return loss requirements. Each transmitter is a single-ended voltage driver. External resistors are used to maintain an accurate source impedance that has a high return loss to the transformer or the capacitor. Each of the drivers can be individually disabled, this is required in fault tolerant applications where redundancy is a requirement. During power-down mode of operation the bipolar outputs can be disabled.

To protect the data integrity during a brownout, the output pulse amplitudes are reduced by a factor of 25% if the supply drops below an internally set limit.

Transmission is possible either with or without a clock. If a clock is used, the transmit input data must consist of full-width NRZ pulses, and the transmitter output pulse width is determined by the duty cycle of the clock. If the transmit clock is tied high, the transmitter output pulses are determined by the input data pulse width. In this mode, RZ data must be supplied to the device.

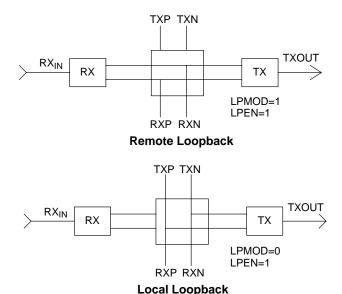


Figure 5. Loopback Configurations





# **Output Transformer Selection**

The 1:1.265 ratio output transformer is recommended for the XR-T5794 because this ratio gives the best possible transmitter output return loss for  $120\Omega$  balanced E1 service. However, other transformers may provide an adequate return loss for many applications. The two characteristics that determine series build-out resistor requirements are:

- Driver output impedance is less than  $5\Omega$ .
- V<sub>s</sub>, which is the driver open circuit output voltage, is 4.5V peak.

The following method may be used to determine transformer suitability for a given use.

1. List the application requirements.

Transformer Ratio = 1:n Vo = Peak Output Pulse Amplitude R<sub>L</sub>= Load Resistance

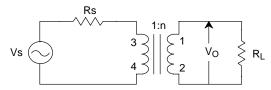


Figure 6. Equivalent Impedance Schematic

2. Calculate equivalent output voltage and load resistance without the transformer.

$$Req = \frac{R_L}{n^2} \qquad Veq = \frac{V_o}{n}$$

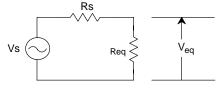


Figure 7. Equivalent Simplified Schematic

3. Calculate the source resistance, R<sub>S</sub>.

$$Rs = Req \left( \frac{Vs}{Veq} - 1 \right)$$

4. Now calculate the theoretical return loss.

Return Loss = 20 
$$log \left( \frac{Req + Rs}{Req - Rs} \right)$$

The calculation given below uses the recommended 1:1.265 ratio transformer as an example:

Transformer Ratio = 1:1.265

$$V_O = 3.0V$$
 Peak

$$R_I = 120\Omega$$

$$Req = \frac{R_L}{n^2} = \frac{120}{1.6} = 75\Omega$$

$$Veq = \frac{V_o}{n} = \frac{3.0}{1.265} = 2.37V$$

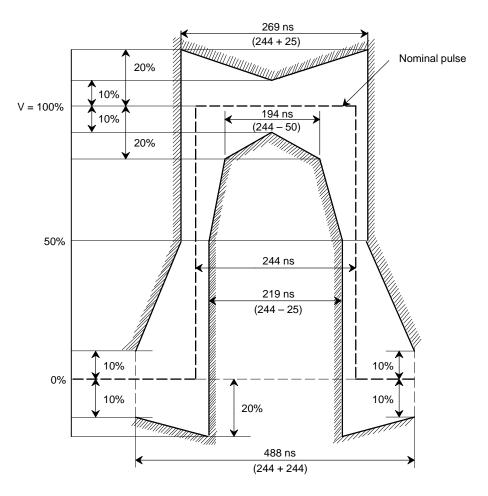
$$Rs = Req\left(\frac{Vs}{Veq} - 1\right) = 75\left(\frac{4.5}{2.37} - 1\right) = 67.4\Omega$$

(Datasheet specifies standard value of  $68\Omega$ )

Calculate the theoretical return loss to determine if the transformer is acceptable.

Return Loss = 
$$20 \log \left( \frac{75 + 67.4}{75 - 67.4} \right) = 25.5 dB$$





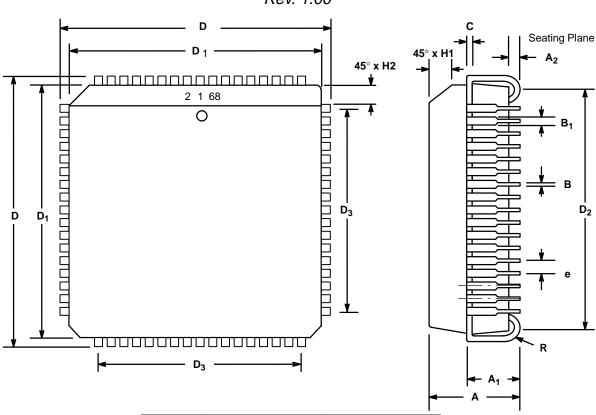
Note: V corresponds to the nominal peak value

Figure 8. CCITT G.703 Pulse Template



# 68 LEAD PLASTIC LEADED CHIP CARRIER (PLCC)

Rev. 1.00



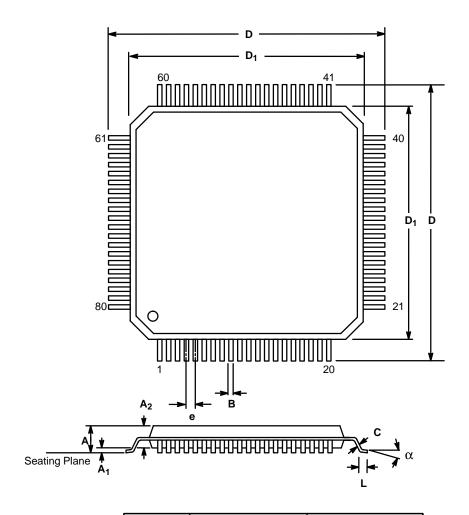
	INC	CHES	MILLI	METERS
SYMBOL	MIN	MAX	MIN	MAX
А	0.165	0.200	4.19	5.08
A <sub>1</sub>	0.090	0.130	2.29	3.30
A <sub>2</sub>	0.020		0.51	
В	0.013	0.021	0.33	0.53
B <sub>1</sub>	0.026	0.032	0.66	0.81
С	0.008	0.013	0.19	0.32
D	0.985	0.995	25.02	25.27
D <sub>1</sub>	0.950	0.958	24.13	24.33
D <sub>2</sub>	0.890	0.930	22.61	23.62
D <sub>3</sub>	0.8	00 typ.	20.3	32 typ.
е	0.050 BSC		1.2	7 BSC
H1	0.042	0.056	1.07	1.42
H2	0.042	0.048	1.07	1.22
R	0.025	0.045	0.64	1.14

Note: The control dimension is the inch column



# 80 LEAD THIN QUAD FLAT PACK (14 x 14 x 1.4 mm, TQFP)

Rev. 3.00



	INCHES		MILLIMETERS	
SYMBOL	MIN	MAX	MIN	MAX
Α	0.055	0.063	1.40	1.60
A <sub>1</sub>	0.002	0.006	0.05	0.15
A <sub>2</sub>	0.053	0.057	1.35	1.45
В	0.009	0.015	0.22	0.38
С	0.004	0.008	0.09	0.20
D	0.622	0.638	15.80	16.20
D <sub>1</sub>	0.547	0.555	13.90	14.10
е	0.0256 BSC		0.65 BSC	
L	0.018	0.030	0.45	0.75
α	0°	7°	0°	7°

Note: The control dimension is the millimeter column





# **Notes**





# **Notes**





# **Notes**





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Datasheet June 1997

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