

Dual 1:2 Low Additive Jitter LVDS Buffer

Check for Samples: CDCLVD2102

FEATURES

- Dual 1:2 Differential Buffer
- Low Additive Jitter <300 fs RMS in 10-kHz to 20-MHz
- Low Within Bank Output Skew of 15 ps (Max)
- Universal Inputs Accept LVDS, LVPECL, LVCMOS
- One Input Dedicated for Two Outputs
- Total of 4 LVDS Outputs, ANSI EIA/TIA-644A Standard Compatible
- Clock Frequency up to 800 MHz
- 2.375–2.625V Device Power Supply
- LVDS Reference Voltage, V_{AC_REF}, Available for Capacitive Coupled Inputs
- Industrial Temperature Range –40°C to 85°C
- Packaged in 3mm × 3mm 16-Pin QFN (RGT)
- ESD Protection Exceeds 3 kV HBM, 1 kV CDM

APPLICATIONS

- Telecommunications/Networking
- Medical Imaging
- Test and Measurement Equipment
- Wireless Communications
- General Purpose Clocking

DESCRIPTION

The CDCLVD2102 clock buffer distributes two clock inputs (IN0, IN1) to a total of 4 pairs of differential LVDS clock outputs (OUT0, OUT3). Each buffer block consists of one input and 2 LVDS outputs. The inputs can either be LVDS, LVPECL, or LVCMOS.

The CDCLVD2102 is specifically designed for driving $50-\Omega$ transmission lines. If driving the inputs in single ended mode, the appropriate bias voltage (V_{AC_REF}) should be applied to the unused negative input pin.

Using the control pin (EN), outputs can be either disabled or enabled. If the EN pin is left open two buffers with all outputs are enabled, if switched to a logical "0" both buffers with all outputs are disabled (static logical "0"), if switched to a logical "1", one buffer with two outputs is disabled and another buffer with two outputs is enabled. The part supports a fail safe function. It incorporates an input hysteresis, which prevents random oscillation of the outputs in absence of an input signal.

The device operates in 2.5V supply environment and is characterized from -40°C to 85°C (ambient temperature). The CDCLVD2102 is packaged in small 16-pin, 3-mm × 3-mm QFN package.

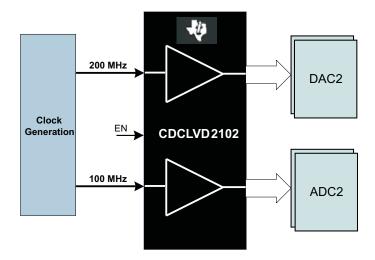


Figure 1. Application Example



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.





These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

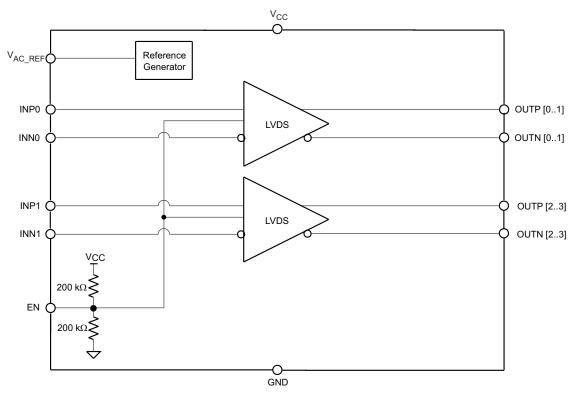
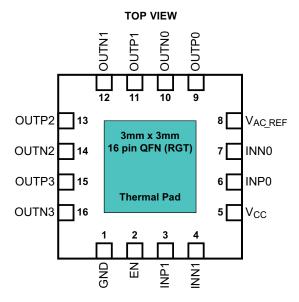


Figure 2. CDCLVD2102 Block Diagram





PIN FUNCTIONS

PIN	PIN TYPE		DES	SCRIPTION				
NAME	NO.							
VCC	5	Power	2.5V supplies for the device					
GND	1	Ground	Device ground					
INP0, INN0	6, 7	Input	Differential input pair or single ended	input				
INP1, INN1	3, 4	Input	Differential redundant input pair or sin	gle ended input				
V _{AC_REF}	8	Output	Bias voltage output for capacitive cou use a 0.1µF to GND on this pin	oled inputs. If used, it is recommended to				
OUTP0, OUTN0	9, 10	Output	Differential LVDS output pair no. 0	INIDO/ININIO in the circust				
OUTP1, OUTN1	11, 12	Output	Differential LVDS output pair no. 1	INP0/INN0 is the input				
OUTP2, OUTN2	13, 14	Output	Differential LVDS output pair no. 2	INIDA/ININIA io the input				
OUTP3, OUTN3	15, 16	Output	Differential LVDS output pair no. 3	INP1/INN1 is the input				
EN	2	Input with internal 200kΩ pull-up and pull-down	Control pin – enables or disables the	Control pin – enables or disables the outputs, (See Table 1)				
Thermal Pad			See thermal management recommend	dations				

Table 1. Output Control Table

EN	CLOCK OUTPUTS					
0	All outputs disabled (static "0")					
OPEN	All outputs enabled					
1	OUT0, OUT1 enabled and OUT2, OUT3 disabled (static "0")					

ABSOLUTE MAXIMUM RATINGS

over operating free-air temperature range (unless otherwise noted) (1)

	VALUE / UNIT
Supply voltage range, V _{CC}	-0.3 to 2.8 V
Input voltage, V _I	-0.2 to (V _{CC} + 0.2) V
Output voltage range, V _O	-0.2 to (V _{CC} + 0.2) V
Driver short circuit current, I _{OSD}	See Note (2)
Electrostatic discharge (HBM, 1.5 kΩ, 100 pF)	>3000 V

⁽¹⁾ Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions beyond those indicated under recommended operating conditions" is not implied. Exposure to absolute—maximum—rated conditions for extended periods may affect device reliability.

RECOMMENDED OPERATING CONDITIONS

	MIN	TYP	MAX	UNITS
Device supply voltage, V _{CC}	2.375	2.5	2.625	V
Ambient temperature, T _A	-40		85	оС

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⁽²⁾ The outputs can handle permanent short.



THERMAL INFORMATION

		CDCLVD2102		
	THERMAL METRIC ⁽¹⁾	RGT	UNITS	
		16 PINS		
θ_{JA}	Junction-to-ambient thermal resistance	51.3		
θ _{JC(top)}	Junction-to-case(top) thermal resistance	85.4		
θ_{JB}	Junction-to-board thermal resistance	20.1	0000	
ΨJΤ	Junction-to-top characterization parameter	1.3	°C/W	
ΨЈВ	Junction-to-board characterization parameter	19.4		
θ _{JC(bottom)}	Junction-to-case(bottom) thermal resistance	6		

⁽¹⁾ For more information about traditional and new thermal metrics, see the IC Package Thermal Metrics application report, SPRA953.

ELECTRICAL CHARACTERISTICS:

At V_{CC} = 2.375 V to 2.625 V and T_A = -40°C to 85°C (unless otherwise noted).

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
EN PIN INP	PUT CHARACTERISTICS				,	
Vd _{I3}	3-State	Open		0.5×V _{CC}		V
Vd _{IH}	Input high voltage		0.7×V _{CC}			V
Vd _{IL}	Input low voltage				0.2×V _{CC}	V
Id _{IH}	Input high current	V _{CC} = 2.625 V, V _{IH} = 2.625 V			30	μΑ
ld _{IL}	Input low current	V _{CC} = 2.625 V, V _{IL} = 0 V			-30	μΑ
R _{pull(EN)}	Input pull-up/ pull-down resistor			200		kΩ
	OS (see Figure 7) INPUT CHARACTER	ISTICS				
f _{IN}	Input frequency				200	MHz
V_{th}	Input threshold voltage	External threshold voltage applied to complementary input	1.1		1.5	V
V _{IH}	Input high voltage		V _{th} + 0.1		V _{CC}	V
V _{IL}	Input low voltage		0		V _{th} - 0.1	V
I _{IH}	Input high current	V _{CC} = 2.625 V, V _{IH} = 2.625 V			10	μΑ
I _{IL}	Input low current	V _{CC} = 2.625 V, V _{IL} = 0 V			-10	μΑ
ΔV/ΔΤ	Input edge rate	20% – 80%	1.5			V/ns
C _{IN}	Input capacitance			2.5		pF
DIFFEREN	TIAL INPUT CHARACTERISTICS					
f _{IN}	Input frequency	Clock input			800	MHz
V _{IN, DIFF}	Differential input voltage peak-to-peak	V _{ICM} = 1.25 V	0.3		1.6	V _{P-P}
V _{ICM}	Input common-mode voltage range	V _{IN, DIFF, PP} > 0.4V	1	\	/ _{CC} – 0.3	V
I _{IH}	Input high current	V _{CC} = 2.625 V, V _{IH} = 2.625 V			10	μΑ
I _{IL}	Input low current	V _{CC} = 2.625 V, V _{IL} = 0 V			-10	μΑ
ΔV/ΔΤ	Input edge rate	20% to 80%	0.75			V/ns
C _{IN}	Input capacitance			2.5		pF

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ELECTRICAL CHARACTERISTICS: (continued)

At V_{CC} = 2.375 V to 2.625 V and T_A = -40°C to 85°C (unless otherwise noted).

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
LVDS OUTF	PUT CHARACTERISTICS					
V _{OD}	Differential output voltage magnitude		250		450	mV
ΔV_{OD}	Change in differential output voltage magnitude	$V_{IN, DIFF, PP} = 0.3 \text{ V,R}_L = 100 \Omega$	-15		15	mV
V _{OC(SS)}	Steady-state common mode output voltage		1.1		1.375	V
$\Delta V_{OC(SS)}$	Steady-state common mode output voltage	$V_{IN, DIFF, PP} = 0.6 \text{ V}, R_L = 100 \Omega$	-15		15	mV
V_{ring}	Output overshoot and undershoot	Percentage of output amplitude V _{OD}			10%	
Vos	Output ac common mode	$V_{IN, DIFF, PP} = 0.6 \text{ V}, R_L = 100 \Omega$		25	70	mV_PP
Ios	Short-circuit output current	$V_{OD} = 0 V$			±24	mA
t _{PD}	Propagation delay	$V_{IN, DIFF, PP} = 0.3 V$		1.5	2.5	ns
t _{SK, PP}	Part-to-part skew				600	ps
t _{SK, O_WB}	Within bank output skew				15	ps
t _{SK,O_BB}	Bank-to-bank output skew	Both inputs are phase aligned			100	ps
t _{SK,P}	Pulse skew(with 50% duty cycle input)	Crossing-point-to-crossing-point distortion	-50		50	ps
t _{RJIT}	Random additive jitter (with 50% duty cycle input)	Edge speed = 0.75V/ns, 10 kHz - 20 MHz			0.3	ps, RMS
t _R /t _F	Output rise/fall time	20% to 80%,100 Ω, 5 pF	50		300	ps
I _{CCSTAT}	Static supply current	Outputs unterminated, f = 0 Hz		27	45	mA
I _{CC100}	Supply current	All outputs enabled, $R_L = 100 \Omega$, $f = 100 \text{ MHz}$		49	77	mA
I _{CC800}	Supply current	All outputs enabled, $R_L = 100 \Omega$, $f = 800 \text{ MHz}$		76	106	mA
V _{AC_REF} CH	ARACTERISTICS					
V _{AC_REF}	Reference output voltage	$V_{CC} = 2.5 \text{ V}, I_{load} = 100 \mu A$	1.1	1.25	1.35	V



Typical Additive Phase Noise Characteristics for 100 MHz Clock

	PARAMETER	MIN	TYP	MAX	UNIT
phn ₁₀₀	Phase noise at 100 Hz offset		-132.9		dBc/Hz
phn _{1k}	Phase noise at 1 kHz offset		-138.8		dBc/Hz
phn _{10k}	Phase noise at 10 kHz offset		-147.4		dBc/Hz
phn _{100k}	Phase noise at 100 kHz offset		-153.6		dBc/Hz
phn _{1M}	Phase noise at 1 MHz offset		-155.2		dBc/Hz
phn _{10M}	Phase noise at 10 MHz offset		-156.2		dBc/Hz
phn _{20M}	Phase noise at 20 MHz offset		-156.6		dBc/Hz
t _{RJIT}	Random additive jitter from 10 kHz to 20 MHz		171		fs, RMS

Typical Additive Phase Noise Characteristics for 737.27 MHz Clock

	PARAMETER	MIN	TYP	MAX	UNIT
phn ₁₀₀	Phase noise at 100 Hz offset		-80.2		dBc/Hz
phn _{1k}	Phase noise at 1 kHz offset		-114.3		dBc/Hz
phn _{10k}	Phase noise at 10 kHz offset		-138		dBc/Hz
phn _{100k}	Phase noise at 100 kHz offset		-143.9		dBc/Hz
phn _{1M}	Phase noise at 1 MHz offset		-145.2		dBc/Hz
phn _{10M}	Phase noise at 10 MHz offset		-146.5		dBc/Hz
phn _{20M}	Phase noise at 20 MHz offset		-146.6		dBc/Hz
t _{RJIT}	Random additive jitter from 10 kHz to 20 MHz		65		fs, RMS

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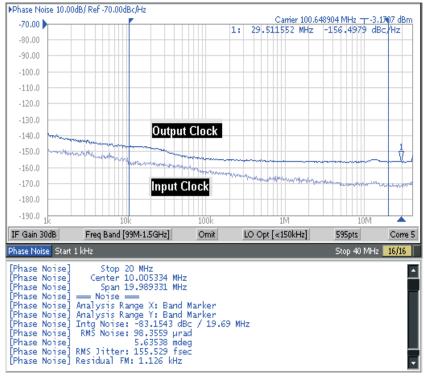


TYPICAL CHARACTERISTICS

INPUT CLOCK AND OUTPUT CLOCK PHASE NOISES

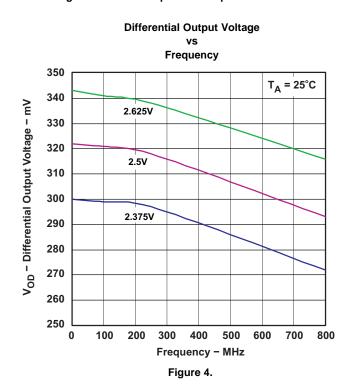
vs

FREQUENCY FROM THE CARRIER ($T_A = 25^{\circ}C$ and $V_{CC} = 2.5V$)



Input clock jitter is 32 fs from 10 kHz to 20 MHz and additive RMS jitter is 152 fs

Figure 3. 100 MHz Input and Output Phase Noise Plot





TEST CONFIGURATIONS

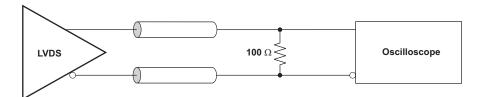


Figure 5. LVDS Output DC Configuration During Device Test

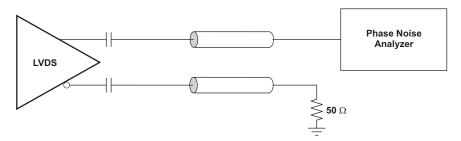


Figure 6. LVDS Output AC Configuration During Device Test

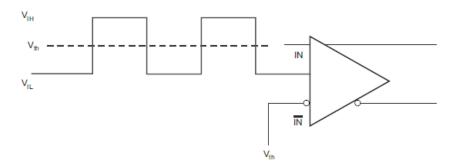


Figure 7. DC Coupled LVCMOS Input During Device Test

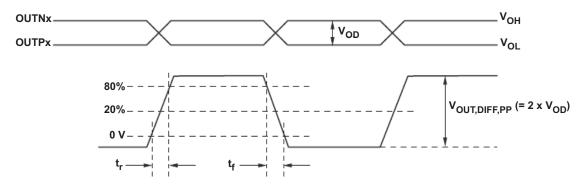
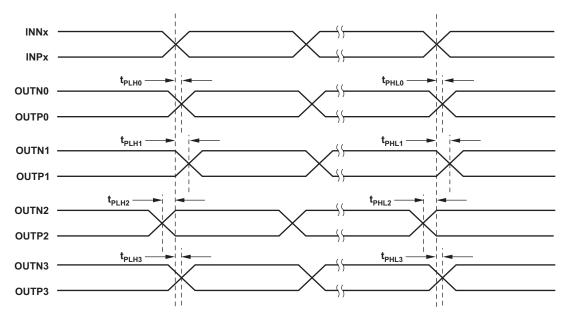


Figure 8. Output Voltage and Rise/Fall Time





- A. Output skew is calculated as the greater of the following: As the difference between the fastest and the slowest t_{PLHn} or the difference between the fastest and the slowest t_{PHLn} (n = 0, 1, 2, 3).
- B. Part-to-part skew is calculated as the greater of the following: As the difference between the fastest and the slowest t_{PLLn} or the difference between the fastest and the slowest t_{PLLn} across multiple devices (n = 0, 1, 2, 3).
- C. Both inputs (IN0 and IN1) are phase aligned.

Figure 9. Output Skew and Part-to-Part Skew

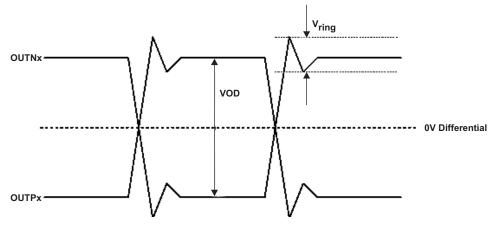


Figure 10. Output Overshoot and Undershoot

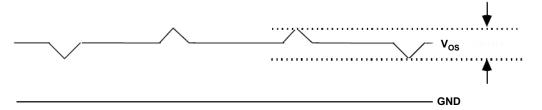


Figure 11. Output AC Common Mode



APPLICATION INFORMATION

THERMAL MANAGEMENT

For reliability and performance reasons, the die temperature should be limited to a maximum of 125°C.

The device package has an exposed pad that provides the primary heat removal path to the printed circuit board (PCB). To maximize the heat dissipation from the package, a thermal landing pattern including multiple vias to a ground plane must be incorporated into the PCB within the footprint of the package. The Thermal Pad must be soldered down to ensure adequate heat conduction to of the package. Figure 12 shows a recommended land and via pattern.

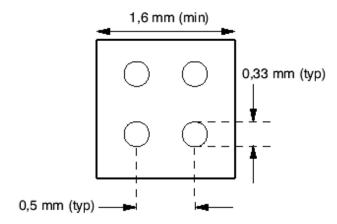


Figure 12. Recommended PCB Layout

POWER-SUPPLY FILTERING

High-performance clock buffers are sensitive to noise on the power supply, which can dramatically increase the additive jitter of the buffer. Thus, it is essential to reduce noise from the system power supply, especially when jitter/phase noise is critical to applications.

Filter capacitors are used to eliminate the low-frequency noise from the power supply, where the bypass capacitors provide the very low impedance path for high-frequency noise and guard the power-supply system against the induced fluctuations. These bypass capacitors also provide instantaneous current surges as required by the device and should have low equivalent series resistance (ESR). To properly use the bypass capacitors, they must be placed very close to the power-supply pins and laid out with short loops to minimize inductance. It is recommended to add as many high-frequency (for example, 0.1 μ F) bypass capacitors as there are supply pins in the package. It is recommended, but not required, to insert a ferrite bead between the board power supply and the chip power supply that isolates the high-frequency switching noises generated by the clock driver; these beads prevent the switching noise from leaking into the board supply. Choose an appropriate ferrite bead with very low dc resistance because it is imperative to provide adequate isolation between the board supply and the chip supply, as well as to maintain a voltage at the supply pins that is greater than the minimum voltage required for proper operation.

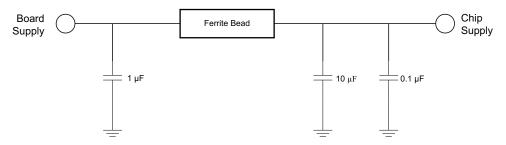


Figure 13. Power-Supply Decoupling

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LVDS OUTPUT TERMINATION

The proper LVDS termination for signal integrity over two 50 Ω lines is 100 Ω between the outputs on the receiver end. Either dc-coupled termination or ac-coupled termination can be used for LVDS outputs. It is recommended to place termination resister close to the receiver. If the receiver is internally biased to a voltage different than the output common mode voltage of the CDCLVD2102, ac-coupling should be used. If the LVDS receiver has internal 100 ohm termination, external termination must be omitted.

Unused outputs can be left open without connecting any trace to the output pins.

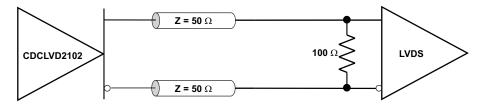


Figure 14. Output DC Termination

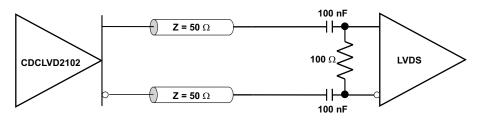


Figure 15. Output AC Termination With Receiver Internally Biased

INPUT TERMINATION

The CDCLVD2102 inputs can be interfaced with LVDS, LVPECL, or LVCMOS drivers.

LVDS Driver can be connected to CDCLVD2102 inputs with dc or ac coupling as shown Figure 16 and Figure 17, respectively.

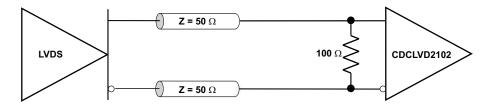


Figure 16. LVDS Clock Driver Connected to CDCLVD2102 Input (DC Coupled)

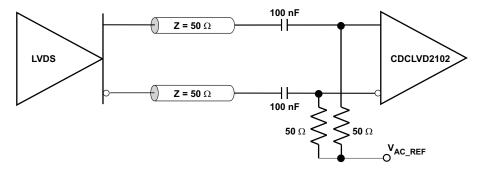


Figure 17. LVDS Clock Driver Connected to CDCLVD2102 Input (AC Coupled)

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Figure 18 shows how to connect LVPECL inputs to the CDCLVD2102. The series resistors are required to reduce the LVPECL signal swing if the signal swing is $>1.6 \text{ V}_{PP}$.

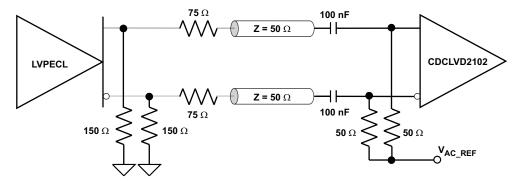


Figure 18. LVPECL Clock Driver Connected to CDCLVD2102 Input

Figure 19 illustrates how to couple a 2.5 V LVCMOS clock input to the CDCLVD2102 directly. The series resistance (R_S) should be placed close to the LVCMOS driver if needed. 3.3 V LVCMOS clock input swing needs to be limited to $V_{IH} \le V_{CC}$.

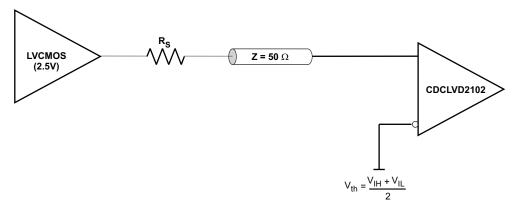


Figure 19. 2.5V LVCMOS Clock Driver Connected to CDCLVD2102 Input

If one of the buffers is used, then the other unused buffer should be disabled through the EN pin, and the unused input pins should be grounded by $1k\Omega$ resistors.



REVISION HISTORY

 Changed Features bullet From: ESD Protection Exceeds 2kV HBM, 500V CDM To: ESD Protection Exceeds HBM, 1 kV CDM 	
Electrostatic discharge was <2000	
• ΔV _{OD} values, MIN was-50, MAX was 50	
V _{OC(SS)} MIN value was 1.125	
• ΔV _{OC(SS)} values, MIN was-50, MAX was 50	
V _{ring} MAX value was 20%	
V _{OS} values, TYP was 30, MAX was 100	
t _{PD} MAX value was 2	5
t _{SK, PP} - deleted the TYP value of 300	ξ
t _R /t _F MIN value was 200	ξ
I _{CCSTAT} MAX value was 42	
Added new paragraph following Figure 19	12



PACKAGE OPTION ADDENDUM

26-Jun-2010

PACKAGING INFORMATION

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Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan ⁽²⁾	Lead/ Ball Finish	MSL Peak Temp ⁽³⁾	Samples (Requires Login)
CDCLVD2102RGTR	ACTIVE	QFN	RGT	16	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	Purchase Samples
CDCLVD2102RGTT	ACTIVE	QFN	RGT	16	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	Purchase Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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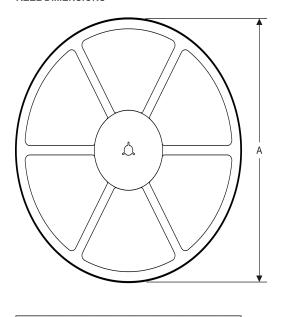
In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

PACKAGE MATERIALS INFORMATION

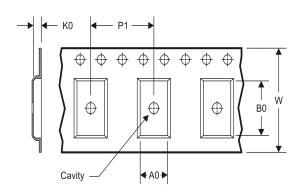
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TAPE AND REEL INFORMATION

REEL DIMENSIONS



TAPE DIMENSIONS



A0	Dimension designed to accommodate the component width
В0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

TAPE AND REEL INFORMATION

*All dimensions are nominal

Device Device	Package Type	Package Drawing		SPQ	Reel Diameter	Reel Width	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
					(mm)	W1 (mm)						
CDCLVD2102RGTR	QFN	RGT	16	3000	330.0	12.4	3.3	3.3	1.1	8.0	12.0	Q2
CDCLVD2102RGTT	QFN	RGT	16	250	330.0	12.4	3.3	3.3	1.1	8.0	12.0	Q2

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*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
CDCLVD2102RGTR	QFN	RGT	16	3000	338.1	338.1	20.6
CDCLVD2102RGTT	QFN	RGT	16	250	338.1	338.1	20.6

RGT (S-PVQFN-N16) PLASTIC QUAD FLATPACK NO-LEAD 3,15 2,85 - A В 3,15 2,85 PIN 1 INDEX AREA TOP AND BOTTOM 0,20 REF. SEATING PLANE 0,08 0,05 0,00 Ċ 16 THERMAL PAD SIZE AND SHAPE SHOWN ON SEPARATE SHEET

NOTES: A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.

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- B. This drawing is subject to change without notice.
- C. Quad Flatpack, No-leads (QFN) package configuration.

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- D. The package thermal pad must be soldered to the board for thermal and mechanical performance.
- E. See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions.

16X $\frac{0,30}{0,18}$

0,50

0,10 M C A B 0,05 M C

4203495/H 10/11

F. Falls within JEDEC MO-220.



RGT (S-PVQFN-N16)

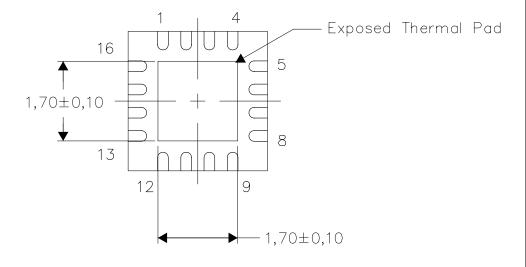
PLASTIC QUAD FLATPACK NO-LEAD

THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No—Lead (QFN) package and its advantages, refer to Application Report, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.



Bottom View

Exposed Thermal Pad Dimensions

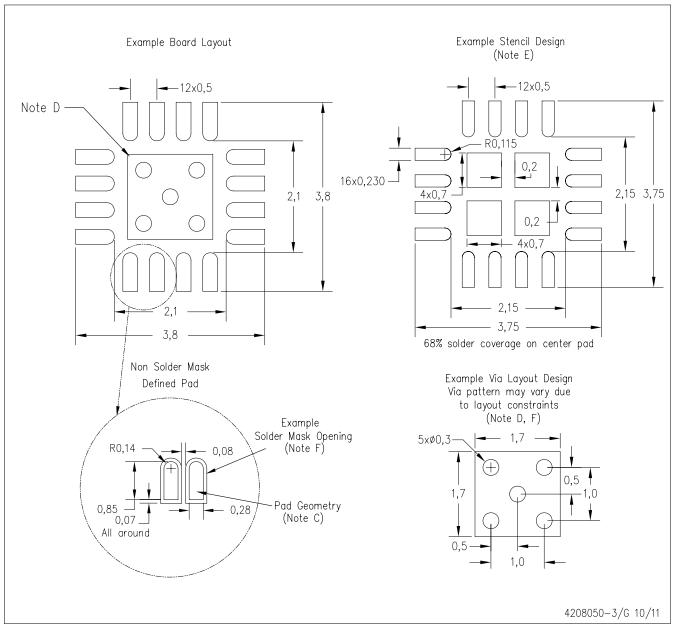
4206349-4/Q 10/11

NOTE: All linear dimensions are in millimeters



RGT (S-PVQFN-N16)

PLASTIC QUAD FLATPACK NO-LEAD



NOTES:

- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com www.ti.com.
- E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
- F. Customers should contact their board fabrication site for minimum solder mask web tolerances between signal pads.



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