## ADS61x9/55xxEVM

## User's Guide

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## 1 Overview

This evaluation module (EVM) user's guide gives an overview of the EVM and provides a general description of the features and functions to be considered while using this module.
This EVM user's guide applies to multiple EVMs:

- ADS61x9 family:
- ADS6128, ADS6148, ADS6129, ADS6149, ADS61B29, ADS61B49
- ADS556x family:
- ADS5560, AD5562
- ADS55xx family:
- ADS5517, ADS5525, ADS5527, ADS5545, ADS5546, ADS5547.


### 1.1 Purpose

The ADS61x9/55xx EVM provides a platform for evaluating the analog-to-digital converter (ADC) under various signal, clock, reference, and power supply conditions. Use this document in combination with the EVM schematic diagram supplied.

### 1.2 EVM Quick-Start Procedure

The ADS61x9/55xx EVM provides numerous options for providing clock, input frequency and power to the ADC under evaluation. The quick start procedure describes how to quickly get initial results using the default configuration of the EVM as it was shipped. The EVM can be put back to default configuration by setting all jumpers the default values as described in Table 1. The default configuration of the EVM is for the Input Frequency (IF) and the clock input is for each to be a single-ended input that is transformer-coupled to the ADC. The default configuration for the power supply is to provide a single 5 V supply to the red banana jack J7, PWR_IN. The default configuration for the EVM is to control the modes of operation by jumper settings for parallel input control pins rather than serial SPI control of the register space. The other modes of operation of the EVM are described in the latter sections of this document.

## CAUTION

Voltage Limits: Exceeding the maximum input voltages can damage EVM components. Undervoltage can cause improper operation of some or all of the EVM components.

A quick-setup procedure for the default configuration of the ADS61x9/55xxEVM follows:.

1. Verify all jumper settings against the schematic jumper list in table 1.

Table 1. Jumper List

| Jumper | Function | Default Jumper Setting |
| :---: | :---: | :---: |
| Interface Circuit Operational Amplifier THS4509 (Bypassed) |  |  |
| SJP1 | AMP_OUT+ | 1-2 |
| SJP2 | AMP_OUT- | 1-2 |
| JP7 | $\overline{\mathrm{PD}}$ | 1-2 |
| SJP5 | AMPIN- | 1-2 |
| ADC Circuit |  |  |
| JP12 | Parallel | 1-2 |
| JP11 | SDA | open |
| JP9 | SEN | 1-2 |
| JP15 | OE | open |
| J2 | DFS | open |
| J3 | MODE | 1-2 |
| J1 | SEN | open |
| Clock Interface Circuit |  |  |
| SJP4 | CLOCKIN | 1-2 |
| SJP7 | CLOCKIN, Y0, Y1P SELECT | 1-2 |
| SJP6 | Y1N SELECT | 1-2 |
| J14 | PWRDWN CDC | 1-2 |
| Power Supply |  |  |
| JP13 | 3.3VA_IN | 1-2 |
| JP14 | 3.3VD_IN | 1-2 |
| JP16 | TPS79501 INPUT SELECT | 1-2 |
| JP19 | 5V_AUX | 1-2 |
| JP17 | TPS5420 INPUT SELECT | NO SHUNT |

2. Connect the $5-\mathrm{V}$ supply between J7 and J12 (GND). If you are using the TSW1200 for capture, it also can be used to source 5 V for the EVM. On the TSW1200, configure JP8 to short 1-2, J22 to short 1-2, and jumper over 5 V from the banana jacks on the TSW1200 to J 7 on the ADC EVM. Do not connect a voltage source greater than 5.5 V .
3. Switch on power supplies.
4. Using a function generator with $50-\Omega$ output impedance, generate a $0-\mathrm{V}$ offset, $1.5-\mathrm{Vpp}$ sine-wave clock into J19. The frequency of the clock must be within the specification for the device speed grade.
5. Use a frequency generator with a $50-\Omega$ output impedance to provide a $0-\mathrm{V}$ offset, $-1-\mathrm{dBFS}$-amplitude sine-wave signal into J6. This provides a transformer-coupled differential input signal to the ADC.
6. Connect the TSW1200 or suitable logic analyzer to J 10 to capture the resulting digital data. If a TSW1200 is being used to capture data, follow the additional alphabetically labeled steps. For more information, see Section 4.
a. After installing the TSW1200 software and connecting the TSW1200 to the USB port, open the TSW1200 software.
b. Depending on the ADC under evaluation, select from the TI ADC Selection pulldown menu.
c. Change the ADC Sample Rate and ADC Input Frequency to match those of the signal generator.
d. After selecting a Single Tone FFT test, press the Capture Data button.

## 2 Circuit Description

### 2.1 Schematic Diagram

The schematic diagram for this EVM is attached at the end of this document. See the schematic or relevant section of this user's guide before changing any jumpers.

### 2.2 Circuit Function

Selection of various modes of operation of the ADS61x9/55xx is most often controlled by jumpers on the EVM, either by placing shunts on 0.025 -inch square jumper posts or by installation of surface mount $0-\Omega$ resistors. In general, the use of $0-\Omega$ resistors as jumpers are used in the clock or signal path where signal integrity is critical and jumper posts are used for static or low-speed control paths. Figure 1] shows the relative location of the jumpers, connectors, and switches used on the ADS61x9/55xx. Figure 2 shows the relative locations of most of the resistors and surface-mount $0-\Omega$ jumper locations used on the EVM. In the description of the circuit options in the following sections, each operational mode is accompanied by a table entry that details the jumper or resistor changes that enable that option. Figure 1] and Figure 2 can assist the user to quickly identify where these jumpers are located on the EVM.


Figure 1. ADS61x9/55xx Jumpers


Figure 2. ADS61x9/55xx Surface Jumpers

The following sections describe the function of individual circuits. See the relevant data sheet for device operating characteristics.

### 2.2.1 Power

Power is supplied to the EVM through banana jacks and from this input power several different ways of delivering power to the ADC and other EVM functions are available. Figure 3 shows a simplified representation of the power options available for the ADS61 $\times 9 / 55 \times x$. The default option is to provide 5 V to the red banana jack J7, and from there the EVM generates 3.3 V for the analog supply to the ADC and 1.8 V for the digital supply to the ADC. The EVM also generates the proper voltages for optional features of the EVM such as the Clock Generation circuitry, the USB circuitry, and the CMOS output buffer.


Figure 3. ADS61x9/55xx Power Distribution
Some ADC devices that may be evaluated on the ADS61x9/55xx platform do not take 1.8 V for the digital supply, but rather require 3.3 V for the digital supply. For this reason an adjustable voltage regulator was chosen to generate the digital supply, and the digital supply may be changed to 3.3 V by changing the value of a resistor, R12. This resistor does not need to be changed in the field unless the ADC is being changed, as the EVM ships with the correct digital supply voltage for the ADC that is installed.
Power for the optional THS4509 operational amplifier is supplied by banana jacks J9 and J11. If the amplifier is being evaluated, 5 V is supplied to J 9 and J 11 is connected to ground. Otherwise, these inputs may be left unconnected.
Although various power options are available on this EVM, care must be taken while applying power on J7 as different options have different voltage ranges specified. Table 2 displays the general jumper setting information; 『able 3 displays the various power option settings. Prior to making any jumper settings, see the schematics located at the end of this document.

Table 2. EVM Power Supply Jumper Description

| EVM Banana Jack | Description | Jumper setting |
| :---: | :--- | :--- |
| J7 | Input | $6-\mathrm{V}$ to $36-\mathrm{V}$ power supply; default - apply just 5 V |
| JP13 | 3.3 VA_IN | $1-2 \rightarrow$ Connect 3.3-V AVDD to TPS79633 output |
| JP14 | 3.3VD_IN | $1-2 \rightarrow$ Connect 3.3-V DVDD to TPS79633 output |
| JP16 | TPS79501 INPUT SELECT | $1-2 \rightarrow$ Connects 5.3 V to IP of TPS79501; 2-3 $\rightarrow$ TPS79501 IP <br> connected to J7 |
| JP19 | 5V_AUX | $2-3 \rightarrow$ TPS79501 op as 5v_Aux rails; $1-2 \rightarrow 5 \mathrm{~V} \_$aux rail <br> connected to J7 |
| JP17 | TS5420 INPUT SELECT | Shunt $\rightarrow$ J7 connected to TPS5420D |

Table 3. EVM Power Supply Options

| EVM Option | Evaluation Goal | Jumper Changes Required | Voltage on J7 | Comments |
| :---: | :---: | :---: | :---: | :---: |
| 1 | Evaluate ADC performance using a cascaded switching power supply (TPS5420D) and LDO solution (TPS79501DCQ) | $\begin{aligned} & \text { JP13 } \rightarrow \text { 1-2; JP14 } \rightarrow \text { 1-2; JP16 } \rightarrow \text { 1-2; } \\ & \text { JP19 } \rightarrow \text { 2-3; JP17 } \rightarrow \text { 1-2; } \end{aligned}$ | $6 \mathrm{~V}-36 \mathrm{~V}$ | Maximum performance and efficiency. |
| 2 | Evaluate ADC performance using a LDO-based solution. | $\begin{aligned} & \text { JP13 } \rightarrow \text { 1-2; JP14 } \rightarrow \text { 1-2; JP16 } \rightarrow \text { 1-2; } \\ & \text { JP19 } \rightarrow \text { 1-2, JP17 } \rightarrow \text { No shunt; } \end{aligned}$ | 5.1 V-5.5 V | Maximum performance. |
| 3 | Evaluate ADC performance using an isolated ADC AVDD and DVDD for current consumption measurements | JP13 $\rightarrow$ connect 3.3 V to pin 2 of Jumper; JP14 $\rightarrow$ connect 3.3 V to pin 2 of Jumper; JP16 $\rightarrow$ No shunt ; JP19 $\rightarrow$ No shunt ; JP17 $\rightarrow$ No shunt; | Do not apply power on J7. | Isolated power supply for current consumption measurements |

### 2.2.1.1 Power Supply Option 1

Option 1 supplies the power to the ADC using cascaded topology of the TPS5420D switching power regulator and the TPS79501DCQ Low Dropout (LDO) regulator. The TPS5420 is a step-down converter which works with the input voltage in the range 6 V to 36 V . The switching supply increases efficiency for higher input voltages but does create noise on the voltage supplies. To reduce the noise, an ultralow-noise, high-PSSR LDO TPS79501DCQ is used to clean the power supply. The TPS5420D is designed for output of 5.3 V , which acts as input for TPS79501. The TPS79501 is designed to output a $5-\mathrm{V}$ output, which is the AVDD for the ADC. This voltage rail is input to the LDO TPS79633, which outputs 3.3 V , used for DVDD for the ADC. A separate TPS79633 is designed to output 3.3 V for the CDCE72010 power supply rail. This solution adds two features to the EVM: one is to increase the range of the power supply on jumper J 7 from 6 V to 36 V , allowing the user to choose any power supply source in the specified range without causing significant power dissipation. The other feature is that the output voltage rail has a much lower ripple, ensuring the better performance of the part even when the power source is fluctuating.

### 2.2.1.2 Power Supply Option 2

Option 2 supplies power to the ADC using the LDOs TPS79633DCQ and TPS79601DCQ. The LDOs limit the power supply on J 7 to be in the range 5.1 V to 5.5 V only. When using this option, take care powering up the EVM as higher voltage or reverse polarity can damage the EVM. This is the default power supply configuration for the ADS61x9/55xx.

### 2.2.1.3 Power Supply Option 3

Option 3 is used to evaluate ADC performance using an isolated AVDD and DVDD power supply for current consumption measurements. This option must be used with caution as reversing the power supply or connecting to the wrong connector can result in damage to the EVM. One common usage of this option is to measure the separate current consumption of the relative supplies under particular operating conditions. For this option, the shunts on jumpers JP13 and JP14 are removed and the input power is supplied to the center post of the jumper. For convenience, a ground post is provided next to the center post for header connections that contain power and ground on 0.1 " centers.

### 2.2.2 Clock Input

The clock can be supplied to the ADC in one of several ways. The default clocking option is to supply a single-ended clock directly to the SMA connecter J19 directly, and this clock is converted to differential and AC coupled to the ADC by transformer coupling. The clock input must be from a clean, low-jitter source and is commonly filtered by a narrow bandpass filter. The clock amplitude is commonly set to about $1.5-\mathrm{V}$ peak-to-peak, and the amplitude offset is not an issue due to the AC coupling of the clock input. The clock source is commonly synchronized with the clock source of the input frequency to keep the clock and IF coherent for meaningful FFT analysis.
Alternatively, the clock may be supplied by an onboard VCXO and CDCE72010 clock buffer. The CDCE72010 Clock Buffer has been factory programmed to output a clock to the ADC that is $1 / 4$ the rate of the onboard VCXO. While using this clock option, a separate $20-\mathrm{MHz}$ reference clock must be supplied to the CDCE72010 by way of the Clock Input SMA connector J19. From the CDCE7201 two clocking options to the ADC are possible. A differential LVPECL clock output may be connected to the ADC clock input or a single-ended CMOS clock from the CDCE72010 may be routed to the ADC transformer-coupled clock input through an onboard crystal filter. For better performance, selecting the CMOS clock through a crystal output is recommended. Prior to making any jumper settings, see the schematic located at the end of this document. Table 5 displays the various clock option settings. The VCXO and crystal filter do not come populated on the EVM by default, although the CDCE72010 Clock buffer is installed.

Texas INSTRUMENTS

Table 4. Clock Input Jumper Description

| EVM Banana <br> Jack | Description | Jumper Setting |
| :---: | :--- | :--- |
| J18 | ENABLE VCXO1 TC0-2111 | $1-2 \rightarrow$ VCXO enabled 2-3 $\rightarrow$ VCXO Disabled |
| J19 | Clock supply |  |
| J14 | CDCE72010 power down | $1-2 \rightarrow$ CDCE72010 is power down; Open $\rightarrow$ CDCE72010 is on |
| J15 | CDCE72010 Reset | $1-2 \rightarrow$ Reset, Open $\rightarrow$ Normal operation. (Default) |$|$| $1-2 \rightarrow$ J19 supplies clock directly to ADC; 2-3 $\rightarrow$ Reference clock for |
| :--- |
| CDCE72010 |

Table 5. EVM Clock Input Options

| EVM Option | Evaluation Goal | Jumper Changes Required | Frequency Input on J19 | CDC Configuration Description | Comments |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | Evaluate ADC performance using a sinusoid clock. | $\begin{aligned} & \text { J18 } \rightarrow \text { 2-3; SJP4 } \rightarrow 1-2 ; \text { SJP7 } \rightarrow \\ & \text { 1-2; SJP6 } \rightarrow 1-2 ; \mathrm{J} 14 \rightarrow 1-2 ; \mathrm{J15} \rightarrow \\ & \text { No shunt; } \end{aligned}$ | ADC's <br> Sampling <br> Frequency | NA | Default |
| 2 | Evaluate ADC performance using a crystal filtered LVCMOS clock derived from CDCE72010 | J18 $\rightarrow$ 1-2; SJP4 $\rightarrow$ 2-3; SJP7 $\rightarrow$ 3-4; SJP6 $\rightarrow$ 1-2; J14 $\rightarrow$ No shunt; $\mathrm{J} 15 \rightarrow$ No shunt; | $\begin{aligned} & \text { 20M for } \\ & \text { VCXO@983.0 } \\ & 4 \mathrm{MHz} \end{aligned}$ | Divide VCXO frequency by 4, output on Y0 | Maximum performance. |
| 3 | Evaluate ADC performance using a differential LVPECL clock | J18 $\rightarrow$ 1-2; SJP4 $\rightarrow$ 2-3; SJP7 $\rightarrow$ 5-6; SJP6 $\rightarrow$ 2-3; J14 $\rightarrow$ No shunt; $\mathrm{J} 15 \rightarrow$ No shunt; | 20M for VCXO@983.0 4 MHz | Divide VCXO frequency by 4 , differential LVPECL Clock output on Y1P and Y1N | Not recommended for most applications |

### 2.2.2.1 Clock Option 1

The Clock Option 1 provides a clock to ADC directly from an external source. For the direct supply of the clock to the ADC, a single-ended square or sinusoidal clock input must be applied to J19. The clock frequency must be within the maximum frequency specified for the ADC. The clock input is converted to a differential signal by a Mini-Circuits ${ }^{T M}$ ADT4-1WT, which has an impedance ratio of 4 , implying that voltage applied on J 19 is stepped up by a factor of 2. ADC performance in this case depends on the clock source quality. This option is also the default configuration on the EVM, when it is shipped from the factory. The test result using this option is shown in Figure 7 .

### 2.2.2.2 Clock Option 2

Option 2 uses the onboard VCXO and CDCE72010 to provide a clock to the ADC. The CDCE72010 is used in SPI mode which uses the internal EEPROM to configure the CDCE72010. The EEPROM is programmed in the factory for a divide-by-4 configuration. The EEPROM configuration is shown in Figure 4. The clock at J19 is the reference clock for CDCE72010. The VCXO frequency can be calculated as Fvcxo = Fout x 4 (Fout is the frequency output U0 and U1). The reference clock for CDCE72010 is calculated from Ref Clock $=($ Fvcxo $\times 125) /(48 \times 128)$. This is the clock-to-M divider. When VCXO of frequency 983.04 MHz is used, the calculation results in a reference clock of 20 MHz ; the clock output on

Y0 pin of CDCE72010 is 245.76 MHz . This clock is filtered using the crystal filter with center frequency of 245.76 MHz . By default, the VCXO and the crystal filter are not populated on the EVM, so that the user can populate the components depending on the end application and sampling rate. This configuration is recommended for applications requiring an onboard clock generation scheme. The test result using this option is shown in Figure 8.


Figure 4. CDCE72010 EEPROM Configuration Block Diagram

### 2.2.2.3 Clock Option 3

Option 3 is used for a differential LVPECL clock. This configuration eliminates the need for a crystal filter. It uses the same EEPROM configuration as Option 2, but in this case, the ADC clock pins are connected to Y1N and Y1P. The jumper setting uses the clock output Y1P and Y1N from CDCE72010, to clock ADC. This configuration is not recommended for SNR critical applications. Notice that the clock frequency does not change. The frequency remains the same as in Clock Option 2. The test result using this option is shown in Figure 9 .

### 2.2.3 Analog Inputs

The EVM can be configured to use either a transformer-coupled input or a TH4509 amplifier input, both from a single-ended source. The SMA connector J6 provides the single-ended analog input to the transformer-coupled input circuit to the ADC. The SMA connector J8 is not installed by default, but can be used to bring a differential input clock to the transformer-coupled input or to bring a single-ended input to the THS4509 input circuit. To set the transformer up for one of these options, the EVM must be configured as per the options listed in table 7. See the schematic located at the end of this document prior to making any jumper changes.

Table 6. Analog Input Jumper description

| EVM Banana <br> Jack | Description | Jumper Setting |
| :---: | :--- | :--- |
| J6 | Analog input single-ended. |  |
| J8 | Analog input, can be used with J6 <br> for differential input | Not populated |
| J9 | Power supply + | Apply 5 V |
| J11 | Power Supply - | Ground. |
| SJP1 | AMP OUT+, +v terminal of T1 | $2-3 \rightarrow$ Amp out+ is selected as the source of input to ADC; $1-2 \rightarrow$ Use <br> Analog input from J6 as signal source to ADC (Use with appropriate SJP5 <br> setting) |
| SJP2 | AMP OUT-,--, -v terminal of T1 | $2-3 \rightarrow$ Amp out+ is selected as the source of input to ADC; $1-2 \rightarrow$ Ground, <br> (Default) Also can be used to connect J8 as a differential input along with <br> SJP1 and SJP3 setting. |
| SJP3 | INPUT -ve select | $1-2 \rightarrow$ J8 supplies the analog signal to ADC; 2-3 $\rightarrow+$ +ve Input to amplifier. <br> DEFAULT is No Shunt |
| SJP5 | INPUT +ve select | $1-2 \rightarrow$ J6 supplies the analog signal to ADC. 2-3 $\rightarrow$-ve Input to amplifier. |
| JP7 | Power down for amplifier <br> THS4509 | $2-3 \rightarrow$ Pulls up the pin (Normal operation or amplifier is ON); 1-2 $\rightarrow$ <br> Grounds the pin (Low-power mode or amplifier is off) |

Table 7. EVM Analog Input Options

| EVM Option | Evaluation Goal | Jumper Changes Required | Voltage on J9 and J11 | Analog Signal to ADC | Comments |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | Evaluate ADC performance using direct input to ADC. | SJP1 $\rightarrow$ 1-2; SJP2 $\rightarrow$ 1-2; SJP3 $\rightarrow$ No shunt; SJP5 $\rightarrow$ 1-2; JP7 $\rightarrow$ 1-2; | Do not connect | From J6 | default |
| 2 | Evaluate ADC performance using input through THS4509 | SJP1 $\rightarrow$ 2-3; SJP2 $\rightarrow$ 2-3; SJP3 $\rightarrow$ No shunt; SJP5 $\rightarrow$ 2-3; JP7 $\rightarrow$ 2-3; Install J8; remove R84; | $\mathrm{J} 9 \rightarrow 5 \mathrm{~V}, \mathrm{~J} 11 \rightarrow$ GND | Signal from J 6 is amplified by THS4509 | Used if input signal requires amplification. |

### 2.2.3.1 Analog Input Option 1

Option 1 supplies the transformer coupled input from J6 to ADC. This configuration is the default on the EVM. The test result using this option is shown in Figure 7. A double-transformer input circuit is used to provide better differential to single-ended conversion than a single transformer can provide. The transformers used are both of a $1: 1$ turns ratio, so termination of the $50-\Omega$ input signal path after the transformers can be two $25-\Omega$ resistors terminated to the Common Mode Voltage (VCM) provided by the ADC.

Following the transformer coupling, surface mount pads are provided for several input circuits. By default, the input circuit is configured as shown in the ADS6149 data sheet under the recommended input circuit for high-bandwidth ( $>100 \mathrm{MHz}$ IF) inputs. However, the recommended low-bandwidth input circuit for the ADS6149 or the recommended input circuit for the ADS5547 can be easily implemented on the surface mount pads provided.

### 2.2.3.2 Analog Input Option 2

Option 2 allows the use of an amplifier to provide input to the ADC. TI has a range of wideband operational amplifiers such as THS4508/09/11/13/20. On this EVM, THS4509 is used as an example to amplify the input from J8. The THS4509 is powered up by applying 5 V to J 9 and GND to J11. A differential power supply may also be used to power up the amplifier if common-mode biasing is an issue for DC-coupled applications. See the THS4509 data sheet (SLOS547). The output of the THS4509 is filtered through a band-pass filter before ADC input. The band-pass filter can be designed depending on the end application. By default, the band-pass filter components are not populated as the filter design
depends on the end application. The TI schematic provides an example of a filter that is designed for the frequency band of 10 MHz to 58 MHz . When using the suggested filter, be sure to consider the proper value for R23 and R24 resistors, as the ADC may impose limits on how large these resistors may be while the amplifier may impose limits on how low an impedance it can drive. A key point when designing a filter is to design it for proper load termination. Care must be taken when supplying the input to the board, and ensure that the source impedance is $50 \Omega$. Results can vary due to mismatching of the various source and termination impedances.

### 2.2.4 Digital Outputs

The LVDS digital outputs can be accessed through the J 10 output connector. A parallel $100-\Omega$ termination resistor must be placed at the receiver to properly terminate each LVDS data pair. These resistors are required if the user wants to analyze the signals on an oscilloscope or a logic analyzer. The ADC performance also can be quickly evaluated using the TSW1200 boards as explained in the next section. The TSW1200 automatically terminates the LVDS outputs once the TSW1200 is connected to J10. Alternatively, the ADS61x9/55xx is supplied with a breakout-board to easily connect the LVDS outputs to a logic analyzer pod. This LVDS breakout-board also properly terminates the LVDS outputs once the breakout board is connected to J 10 .
The ADS6149 and most other ADCs that may be evaluated on this EVM also have an option to output the digitized parallel data in the form of single-ended CMOS. If single-ended CMOS is desired, header post connector J 5 is provided for the CMOS output. In order to use the header J5, a CMOS buffer U7 must be installed in place of a bank of 0-ohm resistors that by default steer the outputs to the LVDS connector J10.

## $3 \quad$ TI ADC SPI Control Interface

This section describes the software features accompanying the EVM kit. The TI ADC SPI control software provides full control of the SPI interface, allowing users to write to any of the ADC registers found in the ADC data sheet. For most ADS6149 (and other ADCs evaluated on this EVM) performance evaluations, users do not need to use the TI SPI control software to get evaluation results. Users only need to use the ADC SPI control software when the desired feature is inaccessible because the ADC is in parallel interface mode.

### 3.1 Installing the ADC SPI Control Software

The ADC SPI control software can be installed on a personal computer by running the setup.exe file located on the CD. This file installs the graphical user interface (GUI) along with the USB drivers needed to communicate to the USB port that resides on the EVM. After the software is installed and the USB cable has been plugged in for the first time, the user is prompted to complete the installation of the USB drivers. When prompted, allow the Windows ${ }^{\text {TM }}$ operating system to search for device drivers and automatically find the TI ADC SPI interface drivers. See Figure 5.

Note: Before plugging in the USB cable for the first time, install the TI ADC SPI software. The software installs the drivers necessary for USB communication.


ADS6149/48/29/28 FREQUENTLY USED REGISTERS



FINE GAIN (dB)


Figure 5. TI ADC SPC Interface Screen

## www.ti.com

### 3.2 Setting Up the EVM for ADC SPI Control

Users who want to use the ADC SPI interface must configure four jumpers for proper control of the SPI bus. By default, the EVM comes with the ADC configured in parallel mode. In order to use the SPI interface to control the ADC modes of operation, users must:

- Move jumper JP12 to short positions 2-3, which places the ADC in serial operation mode.
- Move jumper JP11 to short positions 2-3, which allows the USB circuit to control SDATA.
- Move jumper JP9 to short positions 2-3, which allows the USB circuit to control SEN.


### 3.3 Using the TI ADC SPI Interface Software

Once the software is installed and the USB cable is connected, three primary modes of operating the software are available: SPI Register Writes, SPI Register Write Using a Script File, and ADS6149 Frequently Used Registers.

### 3.3.1 SPI Register Writes

The most basic mode of operation allows full control of writing to individual register addresses. In the top of the interface screen (Figure 5), select the ADS6149 ADC tab from the list of tabs present. Next, type the Address Bytes(s) in hexadecimal (hex) and Data Byte(s) in hex, which can be found in the device data sheet. When you are ready to send this command to the ADC, press Enter on your keyboard or press the enter button below the address and data fields. The logic-analyzer-style graph indicator is updated with the patterns sent to the ADC. The default inputs to both the Address Byte(s) and Data Byte(s) fields are hex inputs as designated by the small $x$ in the control. Users can change the default input style by clicking on the $x$ to binary, decimal, octal, or hex. Multiple register writes can be written simply by changing the contents of the Address Byte(s) and Data Byte(s) field and pressing Enter again.

### 3.3.2 SPI Register Write Using a Script File

For situations where the same multiple registers must be written on a frequent basis, users can easily use a text editor to create a script file containing all ADC register writes. An example script file is located in the \IInstall Directory\Script Files\ADS6145_LVDS_CourseGain.txt. Users who want to take advantage of writing their own script files must start by using the ADS6145_LVDS_CourseGain.txt as a template file. When ready to write the contents of the script file to the ADC, users can press the Load Script button, and they are prompted for the file location of their script file. The commands are sent to the ADC when the user acknowledges the selection of the file.

### 3.3.2.1 ADS6149 Frequently Used Registers

For ease of use, several buttons have been added that allow one-click register writes of commonly used features found in Table 8. These are found in the ADS61x9 tab, as these commands are specific to the ADS6149 ADC only. The software writes to the ADC both the contents of the associated address and data when the button is clicked. When the ADS6149 Reset button is pressed, it issues a software reset to the ADC, and it resets the button values to match the contents inside of the ADC. The graph indicator plots the SPI commands written to the ADC when a button has been depressed.

Table 8. ADS6149 Frequently Used Registers

| Default Value |  |
| :--- | :--- |
| ADS6149 Reset |  |
| 2s Complement/offset Binary controlled by the DFS pin | Straight Binary or 2s complement |
| CMOS/LVDS controlled by the DFS pin | DDR LVDS or CMOS |
| Power Down: OFF | Power Down On |
| No Fine Gain | 0 to 6 dB of gain in 0.5-dB increments |
| INT Reference controlled by the MODE pin | EXT Reference or internal reference |
| High-speed operation (>100 MHz sampling) | Low-Speed operation |
| Test Mode: None | Multiple Options Test |

The SPI control of the ADC register space may at a future date also be controlled by the FPGA that is on the TSW1200 Capture Card. The SPI signals SCLK, SEN, and SDATA can be configured to be driven by a board plugged into the connector J 10 . By default, three $0-\Omega$ resistors are installed to connect these three SPI signals to the USB port that is controlled by the SPI software. These three $0-\Omega$ resistors can be moved to allow the SPI port to be controlled by J10 instead.

## 4 Evaluation

### 4.1 TSW1200 Capture Board

The TSW1200 board can be used to analyze the performance of the EVM. The TSW1200EVM is a circuit board that assists designers in prototyping and evaluating the performance of high-speed ADCs that feature parallel or serialized LVDS outputs. The TSW1200 has the LVDS $100-\Omega$ termination resistor on the input interface for ADC outputs.
To start the TSW1200 software, note the following points.


Figure 6. TSW1200 GUI Introduction

1. Select the ADC type to be used before capturing.
2. For test, select Single Tone FFT plot.
3. For the ADC Sampling Rate, type in the value.
4. Type in the ADC Input Frequency. Auto calculation of the input frequency depends on the FFT record length. As soon as the number is entered, the software calculates the coherent input frequency corresponding to that FFT length. This frequency signal must be supplied through the signal generator.
5. Select the FFT Record Length.
6. Select Capture to obtain the plot
7. The Continuous Capture option is used if the user wants to continuously capture the FFT.

Adjust the input level signal to attain the dBFs of approximately -1 .

### 4.2 Quick-Test Results

The user can make the jumper setting as mentioned in table 1. In this configuration, the EVM uses an external clock source from J19 and a direct input signal J6 to the ADC. This setup uses Power Option 2 (Table 3), Clock Option 1 (Table 5), and Analog Input Option 1 (Table 7), which is the default on the EVM. Figure 7 shows the ADC performance capture using TSW1200 with the input signal of a $100-\mathrm{MHz}$ frequency and clock frequency of 245.76 MHz with ADS6149.


Figure 7. Quick-Setup Test Result.

### 4.3 Test Result With Onboard VCXO and Clock Through Crystal Filter

This test uses the VCXO of frequency 983.04 MHz . This setup uses the Power Option 2 (Table 3), Clock Option 2 (Table 5), and Analog Input Option 1 (Fable 7). For this test, the CDCE72010 crystal filter path was chosen to provide the clock to the ADC. The CDCE72010 provides a single-ended clock through output Y0 (Table 5), which is passed through a crystal filter of center frequency 245.76 MHz . This was the example setup; the VCXO and the crystal filter are not populated on the EVM as the values depend on the end-application sampling rate. The capture result for ADS6149 is as shown in Figure 8.


Figure 8. ADC Performance With Clock Through Onboard VCXO, CDCE72010, and Crystal Filter

### 4.4 Test Result With Onboard VCXO and Differential LVPECL Clock

For the same setup as explained in the previous section, when Clock Option 3 (Table 5) was used, the FFT was captured as shown in Figure 9. The test results with Clock Option 2 are better than with Clock Option 3. That is why Option 2 (clock with crystal filter) is recommended over the differential LVPECL output.


Figure 9. ADC Performance With Clock Through Onboard VCXO, CDCE72010 Configured for Differential LVPECL Output

## 5 Physical Description

This section describes the physical characteristics and printed-circuit board (PCB) layout of the EVM.

### 5.1 PCB Layout

The EVM is constructed on a six-layer, 0.062 -inch-thick, PCB using FR-4 material. The individual layers are shown in Figure 10 through Figure 15. The layout features a common ground plane; however, similar performance can be obtained with careful layout using a split ground plane.


Figure 10. Silkscreen and Top Layer


Figure 11. Top Side


Figure 12. Ground Plane 1


Figure 13. Power Plane 1


Figure 14. Power Plane 2


Figure 15. Ground Plane 2


Figure 16. Bottom Side


Figure 17. Silkscreen and Bottom Layer
5.2 Bill of Materials

| Qty | Reference | Not Installed | Value | Foot Print | Part Number | Manufacturer | Toleran ce | Volt | Watt |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 16 | $\begin{aligned} & \text { C11, C12, } \\ & \text { C32, C85, } \\ & \text { C87-C89, } \\ & \text { C92, } \\ & \text { C96-C103 } \end{aligned}$ |  | 10 nF | 603 | GCM188R71H103KA37D | Panasonic | 10\% | 50 V |  |
| 17 | C13, C47, <br> C50, C53, <br> C56,C61, C62, <br> C70, <br> C72,C78-C81, <br> C83, C135, <br> C145, C146 |  | $0.1 \mu \mathrm{~F}$ | 603 | ECJ-1VB1C104K | Panasonic | 10\% | 16 V |  |
| 5 | $\begin{aligned} & \text { C15, C33, } \\ & \text { C34, C68, C69 } \end{aligned}$ |  | $0.1 \mu \mathrm{~F}$ | 603 | GRM188R71H104KA93D | Murata | 5\% | 50 V |  |
| 1 | C41 |  | $2.2 \mu \mathrm{~F}$ | 805 | GRM21BR71E225KA73L | Murata | 10\% | 25 V |  |
| 1 | C42 |  | 33 pF | 603 | 06031A330FAT2A | AVX | 1\% | 100 V |  |
| 1 | C43 |  | $1 \mu \mathrm{~F}$ | 603 | GRM188R61E105KA12D | Murata | 10\% | 25 V |  |
| 4 | $\begin{aligned} & \text { C45, C48, } \\ & \text { C52, C55 } \end{aligned}$ |  | $33 \mu \mathrm{~F}$ | TANT_B | TPSB336K016R0350 | AVX | 10\% | 16 V |  |
| 8 | C46, C49, C54, C71, C73,C82, C84, C144 |  | $10 \mu \mathrm{~F}$ | 805 | ECJ-2FB1A106K | Panasonic | 10\% | 10 V |  |
| 1 | C51 |  | 33 nF | 603 | 06035C333KAT2A | AVX | 10\% | 50 V |  |
| 1 | C60 |  | 100n | 603 | ECJ-1VB1C104K | Panasonic | 10\% | 16 V |  |
| 1 | C63 |  | $22 \mu \mathrm{~F}$ | 805 | ECJ-2FB0J226M | Panasonic | 20\% | 6.3 V |  |
| 1 | C64 |  | 100 pF | 603 | ECJ-1VC1H101J | Panasonic | 5\% | 50 V |  |
| 2 | C65, C66 |  | $0.1 \mu \mathrm{~F}$ | 402 | ECJ-0EB1A104K | Panasonic | 10\% | 10 V |  |
| 1 | C67 |  | $47 \mu \mathrm{~F}$ | TANT_E | TPSE476M020R0150 | AVX | 10\% | 20 V |  |
| 1 | C148 |  | 3.3 pF | 603 | GRM1885C1H3R3CZ01D | Murata | $\pm 0.25 \mathrm{pF}$ | 50 V |  |
| 0 | C74 | Not Installed | 3.3 pF | 603 | GRM1885C1H3R3CZ01D | $\pm 0$ | +/-25pF | 50 V |  |
| 0 | C75, C77 | Not Installed | 47 pF | 603 | ECJ-1VC1H470J | Panasonic | 5\% | 50 V |  |
| 0 | C76, C104 | Not Installed | 5 pF | 603 | ECJ-1VC1H050C | Panasonic | $\pm 0.25 \mathrm{pF}$ | 50 V |  |
| 1 | C86 |  | $10 \mu \mathrm{~F}$ | TANT_A | T491A106M010AT | Kemet | 20\% | 10 V |  |
| 2 | C90, C91 |  | 27 pF | 603 | GRM1885C2A270JA01D | Murata | 5\% | 100 V |  |
| 1 | C93 |  | $0.01 \mu \mathrm{~F}$ | 603 | C0603C103K1RACTU | Kemet | 10\% | 100 V |  |
| 0 | C94, C95 | Not Installed | 5.6 pF | 603 | GRM1885C1H5R6DZ01D | Murata | $\pm 0.5 \mathrm{pF}$ | 50 V |  |
| 0 | C105, C106 | Not Installed | 150 pF | 603 | GRM1885C1H151JA01D | Murata | 5\% | 50 V |  |
| 7 | $\begin{aligned} & \text { C107, C108, } \\ & \text { C120, C136, } \\ & \text { C138, C140, } \\ & \text { C141 } \end{aligned}$ |  | 10 nF | 201 | ECJ-ZEB1A103K | Panasonic | 10 V | 10 V |  |
| 3 | $\begin{aligned} & \text { C109, C110, } \\ & \text { C119 } \end{aligned}$ |  | 10 nF | 603 | 06035C103KAZ2A | AVX | 10\% | 50 V |  |
| 6 | ```C111, C183-C187``` |  | 1uF | 603 | ECJ-1VB1A105K | Panasonic | 10\% | 10 V |  |
| 4 | $\begin{aligned} & \text { C112, C115, } \\ & \text { C117, C127 } \end{aligned}$ |  | $0.022 \mu \mathrm{~F}$ | 201 | GRM033R60J223KE01D | Panasonic | 10\% | 6.3V |  |
| 5 | $\begin{aligned} & \text { C113, C116, } \\ & \text { C118, C128, } \\ & \text { C130 } \end{aligned}$ |  | 100 pF | 201 | GRM0337U1E101JD01D | Panasonic | 5\% | 25 V |  |
| 2 | C114, C132 |  | 22 pF | 603 | 06033J220GBTTR | AVX | 2\% | 25 V |  |
| 6 | $\begin{aligned} & \text { C121, } \\ & \text { C123-C126, } \\ & \text { C139 } \end{aligned}$ |  | 47 nF | 201 | C0603X5R0J473K | TDK | 10\% | 6.3 V |  |
| 1 | C122 |  | $0.1 \mu \mathrm{~F}$ | 402 | ECJ-0EX1C104K | Panasonic | 10\% | 16 V |  |
| 1 | C129 |  | $0.022 \mu \mathrm{~F}$ | 201 | GRM033R60J223KE01D | AVX Corporation | 10\% | 6.3 V |  |
| 2 | C131, C133 |  | 1 pF | 603 | GQM1885C2A1R0CB01D | Murata | 0.25pF | 100 V |  |
| 2 | C134, C137 |  | $0 \Omega$ | 603 | CRCW06030000ZOEA | DALE | 5\% |  | 1/10W |
| 3 | $\begin{aligned} & \text { C142, C143, } \\ & \text { C147 } \end{aligned}$ |  | 10 nF | 402 | C0402C103K3RACTU | Kemet | 10\% | 25 V |  |
| 1 | C188 |  | $4.7 \mu \mathrm{~F}$ | 1206 | GRM31CF51H475ZA01L | Murata | 20\% | 50 V |  |
| 1 | C189 |  | 470 pF | 603 | GRM1885C1H471JA01D | Murata | 5\% | 50 V |  |
| 3 | C233-C235 |  | $2.2 \mu \mathrm{~F}$ | 1206 | ECJ-HVB1A225K | Panasonic | 10\% | 10 V |  |
| 2 | C413, C414 |  | 15 pF | 402 | ECJ-0EC1H150J | Panasonic | 5\% | 50 V |  |

Physical Description

| Qty | Reference | Not Installed | Value | Foot Print | Part Number | Manufacturer | Toleran ce | Volt | Watt |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | D2 |  | B340A-13-F | DIODE_SM_DO_214AC | B340A-13-F | Diodes Inc |  |  |  |
| 1 | D3 |  | BLUE DIFUSED | DIODE_SM_HSMN_C170 | HSMN-C170 | AVAGO |  |  |  |
| 0 | FLT1 | Not Installed | 245.76 MHz | FILTER_8_SM_150×150 | TF2-Q5EC1 _DNI | Toyocom |  |  |  |
| 12 | $\begin{aligned} & \text { JP7,--JP14, } \\ & \text { JP16, } \\ & \text { JP18,JP19, } \\ & \text { J18 } \end{aligned}$ |  | Jumper_1x3 | HDR_THVT_1x3_100 | 22-28-4030 | Molex |  |  |  |
| 4 | JP15, J15, J14,JP17 |  | HEADER 2POS | JUMPER2 | 22-28-4020 | Molex |  |  |  |
| 3 | J1,J2, J3 |  | HEADER 4x2 | hdr4x2_100ctr | 90131-0124 | Molex |  |  |  |
| 1 | J5 |  | HEADER MALE 20x2 POS 0.100 VERT | CON20x2_100ctr_M_tsw1100_ mate | PBC20DAAN | Sullin |  |  |  |
| 2 | J6, J19 |  | SMA_END_JACK_RND | SMA_SMEL_250x215 | 142-0711-821 | Johnson Components |  |  |  |
| 3 | J7, J9, J11 |  | RED | BANANA_JACK | ST-351A | ALLIED <br> ELECTRONICS |  |  |  |
| 0 | J8, J20, J21 | Not Installed | SMA_END_JACK_RND | SMA_SMEL_250x215 | 142-0711-821 | Johnson Components |  |  |  |
| 1 | J10 |  | CONN_QTH_30x2-D-A | conn_QTH_30x2-D-A | QTH-060-02-F-D-A | Samtec |  |  |  |
| 1 | J12 |  | BLK | BANANA_JACK | ST-351B | ALLIED <br> ELECTRONICS |  |  |  |
| 1 | J17 |  | CONN USB TYP B FEM | conn_usb_typb_fem | 897-43-004-90-000 | Milmax |  |  |  |
| 0 | L6,L7 | Not Installed | 150 nH | 603 | 0603-151J | API Delavan Inc | 5\% |  |  |
| 0 | L8,L24 | Not Installed | 68 | 1206 | EXC-ML32A680U | Panasonic |  |  |  |
| 9 | $\begin{aligned} & \text { L9, L16-L21, } \\ & \text { L23, L13 } \end{aligned}$ |  | 68 | 1206 | EXC-ML32A680U | Panasonic |  |  |  |
| 1 | L10 |  | 1 K at 100 MHz | 805 | BLM21AG102SN1D | Murata |  |  |  |
| 0 | L11, L12 | Not Installed | 47 nH | 603 | PE-0603CD470JTT | Pulse | 5\% |  |  |
| 2 | L14, L15 |  | 180 nH | 603 | ELJ-RER18JF3 | Panasonic | 5\% |  |  |
| 1 | L22 |  | $68 \mu \mathrm{~F}$ | IND_SM_MSS1038 | MSS1038-683ML | Coilcraft |  |  |  |
| 2 | RN1, RN2 |  | 22 | rnet8_16_0603 | 742C163220JPTR | CTS | 5\% |  | 62.5 mW |
| 4 | RN3-RN6 |  | $0 \Omega$ | rnet4_8_0603 | YC164-JR-070RL | Yageo |  |  | 62.5 mW |
| 6 | $\begin{aligned} & \text { R3, R5, R59, } \\ & \text { R60, R63, R64 } \end{aligned}$ |  | 3 K | 603 | ERJ-3EKF3001V | Panasonic | 1\% |  | 1/10W |
| 3 | R4, R58, R62 |  | 2K | 603 | ERJ-3EKF2001V | Panasonic | 1\% |  | 1/10W |
| 5 | $\begin{aligned} & \text { R6, R10, R15, } \\ & \text { R18, R35 } \end{aligned}$ |  | 10K | 603 | ERJ-3EKF1002V | Panasonic | 1\% |  | 1/10W |
| 3 | R7, R69, R70 |  | $0 \Omega$ | 603 | ERJ-3GEY0R00V | Panasonic | 5\% |  | 1/10W |
| 1 | R12 |  | 14 K | 603 | RC0603FR-0714KL | Yageo | 1\% |  | 1/10W |
| 5 | $\begin{aligned} & \text { R13, R19, } \\ & \text { R20, R40, R44 } \end{aligned}$ |  | 100 | 603 | ERJ-3EKF1000V | Panasonic | 1\% |  | 1/10W |
| 1 | R16 |  | 10 | 603 | ERJ-3EKF10R0V | Panasonic | 1\% |  | 1/10W |
| 1 | R17 |  | 30.1 K | 603 | MCR03EZPFX3012 | ROHM | 1\% |  | 1/10W |
| 2 | R21, R26 |  | 4.99 | 603 | CRCW06034R99FNEA | Dale/Vishay | 1\% |  | 1/10W |
| 0 | R22, R25 | Not Installed | 49.9 | 603 | ERJ-3EKF49R9V | Panasonic | 1\% |  | 1/10W |
| 2 | R23, R24 |  | 24.9 | 603 | ERJ-3EKF24R9V | Panasonic | 1\% |  | 1/10W |
| 1 | R27 |  | 1K | 603 | MCR03EZPFX1001 | ROHM | 1\% |  | 1/10W |
| 1 | R28 |  | 7.15K | 603 | RT0603DRD077K15L | Yageo | 0.50\% |  | 1/10W |
| 11 | $\begin{aligned} & \text { R29, R39, } \\ & \text { R41, R45, } \\ & \text { R57,R61, R71, } \\ & \text { R77, R78, } \\ & \text { R92, R98 } \end{aligned}$ |  | 10K | 603 | CRCW060310K0FKEA | DALE | 1\% |  | 1/10W |
| 0 | R30, R68, R81 | Not Installed | 49.9 | 603 | ERJ-3EKF49R9V_DNI | Panasonic | 1\% |  | 1/10W |
| 0 | R31, R37 | Not Installed | 121 | 603 | ERJ-3EKF1210V_DNI | Panasonic | 1\% |  | 1/10W |
| 4 | $\begin{aligned} & \text { R32, R43, } \\ & \text { R88, R91 } \end{aligned}$ |  | 130 | 603 | CRCW0603130RFKEA | DALE | 1\% |  | 1/10W |
| 4 | $\begin{aligned} & \text { R33, R80, } \\ & \text { R83, R90 } \end{aligned}$ |  | 82 | 603 | CRCW060382R0FKEA | DALE | 1\% |  | 1/10W |
| 3 | $\begin{aligned} & \text { R34, R100, } \\ & \text { R101 } \end{aligned}$ |  | 49.9 | 603 | ERJ-3EKF49R9V | Panasonic | 1\% |  | 1/10W |
| 2 | R36, R48 |  | 348 | 603 | ERJ-3EKF3480V | Panasonic | 1\% |  | 1/10W |
| 0 | R38, R46 | Not Installed | 68 | 603 | MCR03EZPFX68R0 | ROHM | 1\% |  | 1/10W |
| 1 | R42 |  | 750 | 603 | CRCW0603750RFKEA | DALE |  |  |  |
| 2 | R47, R89 |  | 100 | 603 | CRCW0603100RFKEA | DALE | 1\% |  | 1/10W |

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| Qty | Reference | Not Installed | Value | Foot Print | Part Number | Manufacturer | Toleran ce | Volt | Watt |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | R49 |  | 10K | 603 | ERJ-3GEYJ103V | Panasonic | 5\% |  | 1/10W |
| 1 | R50 |  | 2.21K | 603 | ERJ-3EKF2211V | Panasonic | 1\% |  | 1/10W |
| 1 | R51 |  | 4.7K | 603 | ERJ-3EKF4R71V | Panasonic | 1\% |  | 1/10W |
| 0 | R52 | Not Installed | 10K | 603 | ERJ-3EKF1002V | Panasonic | 1\% |  | 1/10W |
| 1 | R53 |  | 1.5K | 603 | ERJ-3EKF1501V | Panasonic | 5\% |  | 1/10W |
| 0 | R54, R65-R67 | Not Installed | $0 \Omega$ | 603 | ERJ-3GEYOROOV | Panasonic | 5\% |  | 1/10W |
| 2 | R55, R56 |  | 26.7 | 603 | ERJ-3EKF26R7V | Panasonic | 1\% |  | 1/10W |
| 0 | $\begin{array}{\|l\|} \hline \text { R72, R73, } \\ \text { R75, R76, R79 } \end{array}$ | Not Installed | 10K | 603 | CRCW060310K0FKEA | DALE | 1\% |  | 1/10W |
| 0 | R74, R94 | Not Installed | $0 \Omega$ | 603 | CRCW06030000ZOEA | DALE | 5\% |  | 1/10W |
| 0 | R82 | Not Installed | ZERO | 603 | ERJ-3GEYOROOV | Panasonic | 5\% |  | 1/10W |
| 2 | R84, R87 |  | ZERO | 603 | ERJ-3GEYOROOV | Panasonic | 5\% |  | 1/10W |
| 1 | R85 |  | ZERO | 603 | ERJ-3GEYOROOV | Panasonic | 5\% |  | 1/10W |
| 0 | R86 | Not Installed | ZERO | 603 | ERJ-3GEYOR00V | Panasonic | 5\% |  | 1/10W |
| 1 | R93 |  | 3K | 603 | RC0603FR-073KL | Yageo | 1\% |  | 1/10W |
| 1 | R95 |  | 2.2 | 603 | ERJ-3RQF2R2V | Panasoninc | 1\% |  | 1/10W |
| 2 | R96, R97 |  | 30.1K | 603 | ERJ-3EKF3012V | Panasonic | 1\% |  | 1/10W |
| 1 | R99 |  | 93.1K | 603 | ERJ-3EKF9312V | Panasonic | 1\% |  | 1/10W |
| 0 | R102 | Not Installed | 56.2K | 603 | ERJ-3EKF5622V | Panasonic | 1\% |  | 1/10W |
| 1 | R268 |  | 51K | 402 | ERJ-S02F5102X | Panasonic | 1\% |  | 1/10W |
| 6 | $\begin{aligned} & \text { SJP1, SJP2, } \\ & \text { SJP4-SJP6, } \\ & \text { SJP8 } \end{aligned}$ |  | JUMPER_1x3_SMT | smd_bridge_0603 | NO PART | Short 1-2 with 0 ohm resistors |  |  |  |
| 1 | SJP3 |  |  |  | NO PART |  |  |  |  |
| 1 | SJP7 |  | JUMPER_3X2_SMT | smd_bridge_3x2_0603 | NO PART | Short 1-2 with 0 ohm resistors |  |  |  |
| 1 | SW1 |  | SW PUSHBUTTON | SW_RESET_PTS635 | PTS635SL43 | C \& K Switch |  |  |  |
| 2 | SW2, SW3 |  | RESET | $\begin{aligned} & \text { SW_THVT_SPST_PTS635SL4 } \\ & 3 \end{aligned}$ | PTS635SL43 | ITT <br> Industries/C\&K <br> Div |  |  |  |
| 6 | TP1, TP3, TP6, TP7, TP11, TP13 |  | Test Point Black | testpoint | 5001 | Keystone |  |  |  |
| 6 | TP2, TP4, TP5, TP8, TP10, TP12 |  | Test Point White | testpoint | 5002 | Keystone |  |  |  |
| 3 | T1, T2, T5 |  | WBC1-1 | XFMR_WBC4-1W | WBC1-1 | Coilcraft |  |  |  |
| 1 | T4 |  | ADT4-1WT | TRANS_SMVT_CD542_6 | ADT4-1WT | Mini Circuits |  |  |  |
| 1 | U1 |  | ADS61x9/55xx | QFN48 | ADS61x9/55xx | TI |  |  |  |
| 1 | U4 |  | THS4509 | QFN16 | THS4509RGTT | TI |  |  |  |
| 1 | U5 |  | 93C66B | TSSOP8 | 93C66B-I/ST | Microchip |  |  |  |
| 1 | U6 |  | FT245BM | PQFP32 | FT245BM | Future Technology Devices |  |  |  |
| 0 | U7 | Not Installed | SN74AVC16244DGGR | TSSOP_48_496x244_20 | SN74AVC16244DGGR | TI |  |  |  |
| 2 | U8, U14 |  | TPS79633 | SOT_223_6_TG | TPS79633DCQ | TI |  |  |  |
| 1 | U9 |  | TPS79601 | SOT_223_6_TG | TPS79601DCQR | TI |  |  |  |
| 1 | U10 |  | CDCE72010 | QFN64 | CDCE72010RGCT | TI |  |  |  |
| 1 | U11 |  | TPS5420D | SOIC_8_197x157_50 | TPS5420D | TI |  |  |  |
| 1 | U12 |  | TPS79501 | SOT_223_6_TG | TPS79501DCQ | TI |  |  |  |
| 1 | U13 |  | TPS79601 | SOT_223_6_TG | TPS79601DCQ | TI |  |  |  |
| 0 | VCXO1 | Not Installed | TCO-2111T at 983.04 MHz | VCXO_6_CUSTOM2 | $\begin{aligned} & \text { TCO-2111T at } \\ & 983.04 \mathrm{MHZ} \end{aligned}$ | TOYOCOM |  |  |  |
| 1 | Y1 |  | 6.0000 MHz | smd_csm-7_xtal | ECS-60-32-5PDN-TR | ECS |  |  |  |
| 4 |  |  | Screw machine, ph 4-40× 3/8 |  | PMS 4400038 PH | Building Fasteners | PCB legs |  |  |
| 4 |  |  | Stand-off hex .5/4-40THR |  | 1902C | Keystone Electronic |  |  |  |
| 14 |  |  | Shunt connector |  | S9000-ND | DigiKey |  |  |  |

### 5.3 Schematic Drawings

The schematic drawings appear on the following page.





| Title | ADS61X9-55XX Evaluation Module |
| :--- | :--- |






ADS61X9-55XX Evaluation Module | Size | Bocument Number | RDS61X9-55XXEVM-SCH |
| :---: | :---: | :---: |






Default 1.8 V (ADS61 X9)
For 3.3V (ADS55X): $\begin{aligned} & \text { R12 }=51 \mathrm{~K} \\ & \mathrm{C} 42=16 \mathrm{FF}\end{aligned}$


Tite $\quad$ ADS61X9-55XX Evaluation Module

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Exceeding the specified input range may cause unexpected operation and/or irreversible damage to the EVM. If there are questions concerning the input range, please contact a TI field representative prior to connecting the input power.
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During normal operation, some circuit components may have case temperatures greater than $50^{\circ} \mathrm{C}$. The EVM is designed to operate properly with certain components above $25^{\circ} \mathrm{C}$ as long as the input and output ranges are maintained. These components include but are not limited to linear regulators, switching transistors, pass transistors, and current sense resistors. These types of devices can be identified using the EVM schematic located in the EVM User's Guide. When placing measurement probes near these devices during operation, please be aware that these devices may be very warm to the touch.

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