

Product Features

- High dynamic range downconverter with integrated LO, IF, & RF amps
- RF: 1900 2200 MHz
- IF: 65 300 MHz
- +38 dBm Output IP3
- +21 dBm Output P1dB
- 5.3 dB Noise Figure
- +5V Single supply operation
- Pb-free 6mm 28-pin QFN package
- Low-side LO configuration
- Common footprint with other PCS/ cellular versions

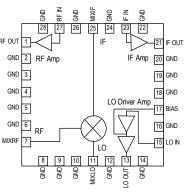
Product Description

The CV111-3A is a high linearity downconverter designed to meet the demanding issues for performance, functionality, and cost goals of current and next generation mobile infrastructure basestations. It provides high dynamic range performance in a low profile surface-mount leadless package that measures 6 x 6 mm square.

Functionality includes RF amplification, frequency conversion and IF amplification, while an integrated LO driver amplifier powers the passive mixer. The MCM is implemented with reliable and mature GaAs MESFET and InGaP HBT technology.

Typical applications include frequency down conversion, modulation and demodulation for receivers used in CDMA, CDMA2000, W-CDMA / IMT2000, GPRS, and EDGE 2.5G mobile infrastructure technologies for UMTS frequency bands.

Functional Diagram



Top View

Specifications (1)

Parameters	Units	Min	Тур	Max	Comments
RF Frequency Range	MHz		1900 – 2200		
LO Frequency Range	MHz		1600 - 2135		
IF Center Frequency Range	MHz		65 - 300		See note 2
% Bandwidth around IF center frequency	%		±7.5		See note 3
IF Test Frequency	MHz		240		
SSB Conversion Gain	dB		20		Temp = 25 °C
Gain Drift over Temp (-40 to 85 °C)	dB		±0.5		Referenced to +25 °C
Output IP3	dBm		+38		See note 4
Output IP2	dBm		+48		See note 4
Output 1dB Compression Point	dBm		+21		
Noise Figure	dB		5.3		See note 5
LO Input Drive Level	dBm	-2.5	0	+2.5	
LO-RF Isolation	dB		40		See note 6
LO-IF Isolation	dB		25		$P_{LO} = 0 \text{ dBm}$
Return Loss: RF Port	dB		14		
Return Loss: LO Port	dB		14		
Return Loss: IF Port	dB		14		
Operating Supply Voltage	V	+4.9	+5	+5.1	
Supply Current	mA	290	360	480	
FIT Rating	failures/1E9 hrs			72.1	@ 70° C ambient, 90% confidence
Thermal Resistance	°C / W			27	
Junction Temperature	°C			160	See note 7

Absolute Maximum Rating

Parameter	Rating
Operating Case Temperature	-40 to +85° C
Storage Temperature	-55 to +125° C
DC Voltage	+6 V
Junction Temperature	+220 °C
RF Input (continuous)	+2 dBm

Ordering Information

Part No.	Description
CV111-3AF	UMTS-band High Linearity Downconverter (lead-free/RoHS-compliant 6x6mm QFN package)
CV111-3APCB75	Fully Assembled Eval. Board, IF = 75MHz
CV111-3APCB240	Fully Assembled Eval. Board, IF = 240MHz

Specifications when using the application specific circuit (shown on page 3) with a low side LO = 0 dBm in a downconverting application over the operating case temperature range. IF matching components affect the center IF frequency. Proper component values for other IF center frequencies than shown can be provided by emailing to applications.engineering@wj.com. The IF bandwidth of the converter is defined as 15% around any center frequency in its operating IF frequency range. The bandwidth is determined with external components. Specifications are valid around the total 47.5% bandwidth, i.e. with a center frequency of 80 MHz, the specifications are valid from 80 ± 6 MHz.

Assumes the supply voltage = +5 V. OIP3 is measured with $\Delta f = 1$ MHz with IF_{out} = 5 dBm / tone.

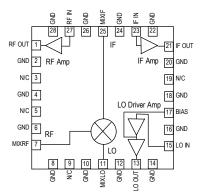
Assumes LO injection noise is filtered at the thermal noise floor, -174 dBm/Hz, at the RF, IF, and Image frequencies.

L-R Isolation is referenced to an LO injection of 0 dBm. The -1.4P performance shown also includes the isolation due to an external SAW filter between the RF amplifier and mixer. The maximum junction temperature ensures a minimum MTTF rating of 1 million hours of usage.

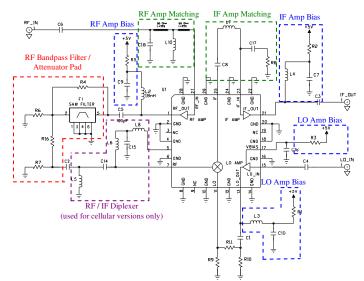
Operation of this device above any of these parameters may cause permanent damage



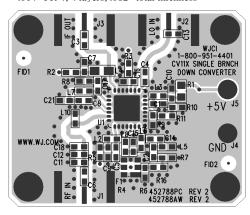
Device Architecture / Application Circuit Information



Typical Downconverter Ferjormance Chain Analysis									
		Output	Outnut			Cumulative Performance			
Stage	Gain (dB)	P1dB (dBm)	Output IP3 (dBm)	NF (dB)	Current (mA)	Gain (dB)	Output P1dB (dBm)	Output IP3 (dBm)	NF (dB)
RF Amplifier	12	21	41	3.5	140	12	21.0	41.0	3.5
RF Filter	-2			2.0		10	19.0	39.0	3.6
LO Amp / MMIC Mixer	-8.5	8	23	9.1	80	1.5	6.1	22.2	4.8
IF Amplifier	18.5	23	41	2.1	140	20	20.7	37.9	5.4
CV111_3A	Cumulativa Parformanca			360	20	20.7	37.0	5.4	



Printed Circuit Board Material: .014" FR-4, 4 layers, .062" total thickness



CV111-3A: The application circuit can be broken up into four main functions as denoted in the colored dotted areas above: RF/IF diplexing (purple; this is only used with the cellular-band CV products), amplifier matching (green), filtering (red), and dc biasing (blue). There are various placeholders for chip components in the circuit schematic so that a common PCB can be used for all WJ single-branch converters. Additional placeholders for other optional functions such as filtering are also included.

RF / IF Amplifier Matching: The RF amplifier requires a shunt matching element for optimal gain and input return loss performance. The IF amplifier requires matching elements to optimize the performance of the amplifier to the desired IF center frequency. Since IF bandwidths are typically on the order of 5 to 10%, a simple two element matching network, in the form of either a high-pass or low-pass filter structure, is sufficient to match the MMIC IF amplifier over these narrow bandwidths. Proper component values for other IF center frequencies can be found in the IF Amplifier Matching Table or by e-mailing to sjcapplication.engineering@tqs.com.

RF Bandpass Filtering: Bandpass filtering is recommended to achieve the best noise figure performance with the downconverter. The bandpass filter, implemented with a SAW filter on the application circuit, allows for the suppression of noise from the image frequency. It is permissible to not use a filter and use a 2 dB pad with R6, R7, and R16 instead with slightly degraded noise figure performance.

External Diplexer: This is only used with the cellular-band CV products. The mixer performs the diplexing internally for the CV111-3A; therefore the components shown in the diplexer section should be loaded as follows: $C2 = C14 = 0 \Omega$.

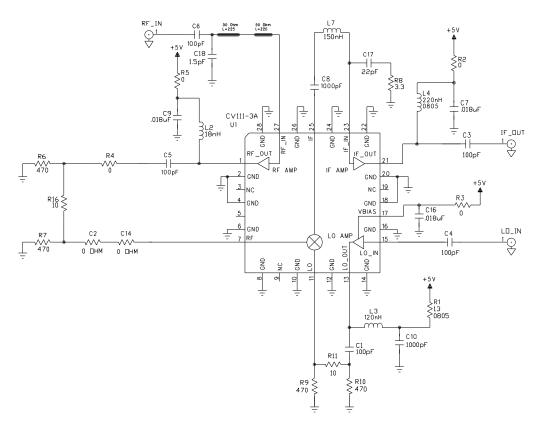
DC biasing: DC bias must be provided for the RF, LO and IF amplifiers in the converter. R1 sets the operating current for the last stage of the LO amplifier and is chosen to optimize the mixer LO drive level. Proper RF chokes and bypass capacitors are chosen for proper amplifier biasing at the intended frequency of operation. The "+5 V" dc bias should be supplied directly from a voltage regulator.

IF Amplifier Matching

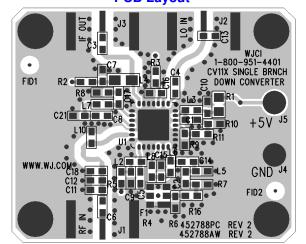
Frequency (MHz)	40	50	75	100	125	130	155	169	180	210	240
L7 (nH)	470	430	150	150	120	120	100	82	82	82	56
C17 (pF)	24	15	22	10	8.2	6.8	5.6	5.0	4.7	3.3	3.9
R8 (ohms)	4.7	4.7	3.3	2.2	2.2	2.2	2.2	2.2	2.2	2.2	2.2
L4 (nH)	470	240	330	330	330	330	330	330	330	220	220



Downconverting Application Circuit: CV111-3APCB75 RF = 1900 - 2200 MHz, IF = 75 MHz



PCB Layout



Circuit Board Material: .014" FR-4, 4 layers, .062" total thickness

Bill of Materials

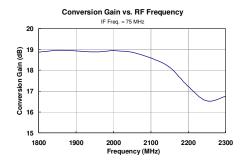
Diff of Matter Mis						
Ref. Desig.	Component					
R1	13 Ω chip resistor, size 0805					
R2, R3, R4, R5, C2, C14	0 Ω chip resistor					
R6, R7, R9, R10	470 Ω chip resistor					
R8	3.3 Ω chip resistor					
R11, R16	10 Ω chip resistor					
C1, C3, C4, C5, C6	100 pF chip capacitor					
C7, C9, C16	0.018 μF chip capacitor					
C8, C10	1000 pF chip capacitor					
C11, C12, C13, C15, C21, F1, L5, L6, L8, L10	Shown in silkscreen, but not used in actual circuit.					
C17	22 pF chip capacitor					
C18	1.5 pF chip capacitor					
L2	18 nH chip inductor					
L3	120 nH chip inductor					
L4	220 nH chip inductor, size 0805					
L7	22 nH chip inductor					
U1	CV111-3A WJ Converter					

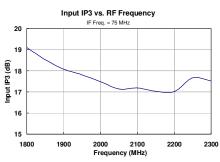
All components are of size 0603 unless otherwise specified.

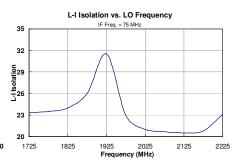


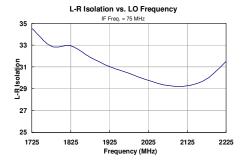


CV111-3APCB75 Application Circuit Performance Plots RF = 1900 - 2200 MHz, IF = 75 MHz



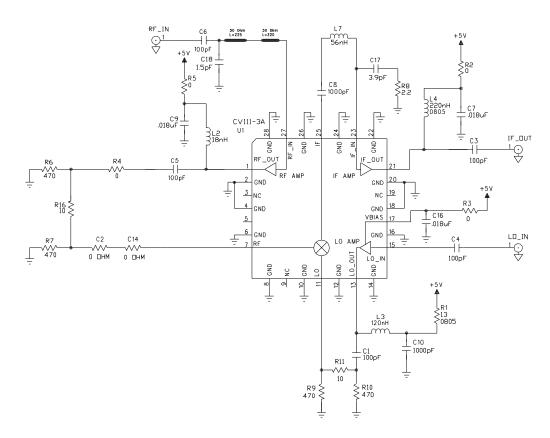




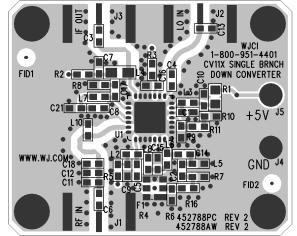




Downconverting Application Circuit: CV111-3APCB240 RF = 1900 - 2200 MHz, IF = 240 MHz



PCB Layout



Circuit Board Material: .014" FR-4, 4 layers, .062" total thickness

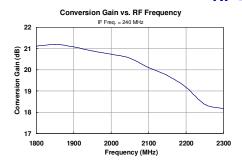
Bill of Materials

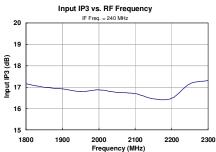
ZIII GI IVIACCI MID						
Ref. Desig.	Component					
R1	13 Ω chip resistor, size 0805					
R2, R3, R4, R5, C2, C14	0 Ω chip resistor					
R6, R7, R9, R10	470 Ω chip resistor					
R8	2.2 Ω chip resistor					
R11, R16	10 Ω chip resistor					
C1, C3, C4, C5, C6	100 pF chip capacitor					
C7, C9, C16	0.018 μF chip capacitor					
C8, C10	1000 pF chip capacitor					
C11, C12, C13, C15, C21, F1, L5, L6, L8, L10	Shown in silkscreen, but not used in actual circuit.					
C17	3.9 pF chip capacitor					
C18	1.5 pF chip capacitor					
L2	18 nH chip inductor					
L3	120 nH chip inductor					
L4	220 nH chip inductor, size 0805					
L7	56 nH chip inductor					
U1	CV111-3A WJ Converter					

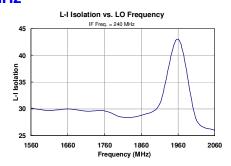
All components are of size 0603 unless otherwise specified.

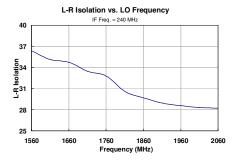


CV111-3APCB240 Application Circuit Performance Plots RF = 1900 - 2200 MHz, IF = 240 MHz







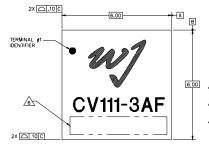




Mechanical Information

This package is lead-free/RoHS-compliant. The plating material on the pins is annealed matte tin over copper. It is compatible with both lead-free (maximum 260 °C reflow temperature) and leaded (maximum 245 °C reflow temperature) soldering processes.

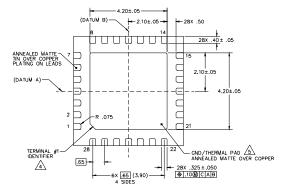
Outline Drawing



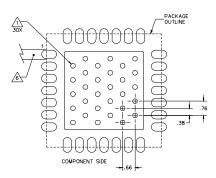
NOTES:

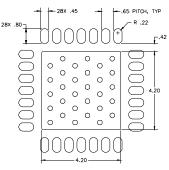
- EXCEPT WHERE NOTED, THIS PART OUTLINE CONFORMS TO JEDEC STANDARD MO-220, ISSUE E (VARIATION VJJC) FOR THERMALLY ENHANCED PLASTIC VERY THIN FINE PITCH QUAD FLAT NO LEAD PACKAGE (QFN).
- DIMENSIONING & TOLERANCING CONFORM TO ASME Y14,4M-1994.
- ALL DIMENSIONS ARE IN MILLIMETERS. ANG
- IN DEGREES.
- THE TERMINAL #1 IDENTIFIER AND TERMINAL NUMBERING CONVENTION CONFORM TO JESD 95-1 SPP-012.
- COPLANARITY APPLIES TO THE EXPOSED GROUND/THERMAL PAD AS WELL AS THE TERMINALS.
- 6 ALPHA-NUMERIC LOT CODE.

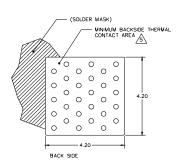




Mounting Configuration / Land Pattern







NOTES:

GROUND/THERMAL VIAS ARE CRITICAL FOR THE PROPER PERFORMANCE OF THIS DEVICE. VIAS SHOULD USE A .35mm (#80/.0135") DIAMETER DRILL AND HAVE A FINAL, PLATED THRU DIAMETER OF .25mm (.010").

- ADD AS MUCH COPPER AS POSSIBLE TO INNER AND OUTER LAYERS NEAR THE PART TO ENSURE OPTIMAL THERMAL PERFORMANCE
- 3. TO ENSURE RELIABLE OPERATION, DEVICE GROUND PADDLE-TO-
- 4. ADD MOUNTING SCREWS NEAR THE PART TO FASTEN THE BOARD TO A HEATSINK. ENSURE THAT THE GROUND/THERMAL VIA REGION CONTACTS THE HEATSINK.
- DO NOT PUT SOLDER MASK ON THE BACK SIDE OF THE F
 BOARD IN THE REGION WHERE THE BOARD CONTACTS THE

6 RF TRACE WIDTH DEPENDS UPON THE PC BOARD MATERIAL AND CONSTRUCTION.

- 7. USE 1 OZ. COPPER MINIMUM
- 8. ALL DIMENSIONS ARE IN MILLIMETERS. ANGLES ARE IN DEGREES.

Product Marking

The component will be lasermarked with a "CV111-3AF" product label with an alphanumeric lot code on the top surface of the package.

Tape and reel specifications for this part will be located on the website in the "Application Notes" section.

ESD / MSL Information



Caution! ESD sensitive device.

ESD Rating: Class 1B

Value: Passes ≥ 500V to <1000V
Test: Human Body Model (HBM)
Standard: JEDEC Standard JESD22-A114

ESD Rating: Class III

Value: Passes ≥ 500V to <1000V
Test: Charged Device Model (CDM)
Standard: JEDEC Standard JESD22-C101

MSL Rating: Level 2 at +260 °C convection reflow Standard: JEDEC Standard J-STD-020

Functional Pin Layout

Pin	Function	Pin	Function		
1	RF Amp Output	15	LO Amp Input		
2	GND	16	GND		
3	N/C or GND	17	LO Amp Bias		
4	GND	18	GND		
5	N/C or GND	19	N/C or GND		
6	GND	20	GND		
7	Mixer RF Input	21	IF Amp Output/Bias		
8	GND	22	GND		
9	N/C or GND	23	IF Amp Input		
10	GND	24	GND		
11	Mixer LO Input	25	Mixer IF Output		
12	GND	26	GND		
13	LO Amp Output/Bias	27	RF Amp Input		
14	GND	28	GND		

Specifications and information are subject to change without notice